

SN74SSTU32866

25-BIT CONFIGURABLE REGISTERED BUFFER WITH ADDRESS-PARITY TEST

SCES564A – APRIL 2004 – REVISED NOVEMBER 2007

- Member of the Texas Instruments Widebus+™ Family
- Pinout Optimizes DDR2 DIMM PCB Layout
- Configurable as 25-Bit 1:1 or 14-Bit 1:2 Registered Buffer
- Chip-Select Inputs Gate the Data Outputs from Changing State and Minimizes System Power Consumption
- Output Edge-Control Circuitry Minimizes Switching Noise in an Unterminated Line
- Supports SSTL_18 Data Inputs
- Differential Clock (CLK and $\overline{\text{CLK}}$) Inputs
- Supports LVCMOS Switching Levels on the Control and $\overline{\text{RESET}}$ Inputs
- Checks Parity on DIMM-Independent Data Inputs
- Able to Cascade with a Second SN74SSTU32866
- $\overline{\text{RESET}}$ Input Disables Differential Input Receivers, Resets All Registers, and Forces All Outputs Low, Except $\overline{\text{QERR}}$
- Latch-Up Performance Exceeds 100 mA Per JESD 78, Class II
- ESD Protection Exceeds JESD 22
 - 2000-V Human-Body Model (A114-A)
 - 200-V Machine Model (A115-A)
 - 1000-V Charged-Device Model (C101)

description/ordering information

This 25-bit 1:1 or 14-bit 1:2 configurable registered buffer is designed for 1.7-V to 1.9-V V_{CC} operation. In the 1:1 pinout configuration, only one device per DIMM is required to drive nine SDRAM loads. In the 1:2 pinout configuration, two devices per DIMM are required to drive 18 SDRAM loads.

All inputs are SSTL_18, except the reset ($\overline{\text{RESET}}$) and control (Cn) inputs, which are LVCMOS. All outputs are edge-controlled circuits optimized for unterminated DIMM loads and meet SSTL_18 specifications, except the open-drain error ($\overline{\text{QERR}}$) output.

The SN74SSTU32866 operates from a differential clock (CLK and $\overline{\text{CLK}}$). Data are registered at the crossing of CLK going high and $\overline{\text{CLK}}$ going low.

The SN74SSTU32866 accepts a parity bit from the memory controller on the parity bit (PAR_IN) input, compares it with the data received on the DIMM-independent D-inputs (D2–D3, D5–D6, D8–D25 when C0 = 0 and C1 = 0; D2–D3, D5–D6, D8–D14 when C0 = 0 and C1=1; or D1–D6, D8–D13 when C0 = 1 and C1=1) and indicates whether a parity error has occurred on the open-drain $\overline{\text{QERR}}$ pin (active low). The convention is even parity; i.e., valid parity is defined as an even number of ones across the DIMM-independent data inputs, combined with the parity input bit. To calculate parity, all DIMM-independent data inputs must be tied to a known logic state.

When used as a single device, the C0 and C1 inputs are tied low. In this configuration, parity is checked on the PAR_IN input signal, which arrives one cycle after the input data to which it applies. Two clock cycles after the data are registered, the corresponding partial-parity-out (PPO) and $\overline{\text{QERR}}$ signals are generated.

ORDERING INFORMATION

T _A	PACKAGE†	ORDERABLE PART NUMBER	TOP-SIDE MARKING
0°C to 70°C	LFBGA – GKE Tape and reel	SN74SSTU32866GKER	SU866

† Package drawings, standard packing quantities, thermal data, symbolization, and PCB design guidelines are available at www.ti.com/sc/package.



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description/ordering information (continued)

When used in pairs, the C0 input of the first register is tied low, and the C0 input of the second register is tied high. The C1 input of both registers are tied high. Parity, which arrives one cycle after the data input to which it applies, is checked on the PAR_IN input signal of the first device. Two clock cycles after the data are registered, the corresponding PPO and \overline{QERR} signals are generated on the second device. The PPO output of the first register is cascaded to the PAR_IN of the second SN74SSTU32866. The \overline{QERR} output of the first SN74SSTU32866 is left floating, and the valid error information is latched on the \overline{QERR} output of the second SN74SSTU32866.

If an error occurs and the \overline{QERR} output is driven low, it stays latched low for a minimum of two clock cycles or until \overline{RESET} is driven low. If two or more consecutive parity errors occur, the \overline{QERR} output is driven low and latched low for a clock duration equal to the parity-error duration or until \overline{RESET} is driven low. The DIMM-dependent signals (\overline{DCKE} , \overline{DCS} , \overline{DODT} , and \overline{CSR}) are not included in the parity-check computation.

The C0 input controls the pinout configuration of the 1:2 pinout from register-A configuration (when low) to register-B configuration (when high). The C1 input controls the pinout configuration from 25-bit 1:1 (when low) to 14-bit 1:2 (when high). C0 and C1 should not be switched during normal operation. They should be hard-wired to a valid low or high level to configure the register in the desired mode. In the 25-bit 1:1 pinout configuration, the A6, D6, and H6 terminals are driven low and are do-not-use (DNU) pins.

In the DDR2 RDIMM application, \overline{RESET} is specified to be completely asynchronous with respect to CLK and \overline{CLK} . Therefore, no timing relationship can be ensured between the two. When entering reset, the register is cleared, and the data outputs are driven low quickly, relative to the time required to disable the differential input receivers. However, when coming out of reset, the register becomes active quickly, relative to the time required to enable the differential input receivers. As long as the data inputs are low, and the clock is stable during the time from the low-to-high transition of \overline{RESET} until the input receivers are fully enabled, the design of the SN74SSTU32866 ensures that the outputs remain low, thus ensuring there will be no glitches on the output.

To ensure defined outputs from the register before a stable clock has been supplied, \overline{RESET} must be held in the low state during power up.

The device supports low-power standby operation. When \overline{RESET} is low, the differential input receivers are disabled, and undriven (floating) data, clock, and reference voltage (V_{REF}) inputs are allowed. In addition, when \overline{RESET} is low, all registers are reset and all outputs are forced low, except \overline{QERR} . The LVCMOS \overline{RESET} and Cn inputs always must be held at a valid logic high or low level.

The device also supports low-power active operation by monitoring both system chip select (\overline{DCS} and \overline{CSR}) inputs and gates the Qn and PPO outputs from changing states when both \overline{DCS} and \overline{CSR} inputs are high. If either \overline{DCS} or \overline{CSR} input is low, the Qn and PPO outputs function normally. Also, if the internal low-power signal ($\overline{LPS1}$) is high (one cycle after \overline{DCS} and \overline{CSR} go high), the device gates the \overline{QERR} output from changing states. If $\overline{LPS1}$ is low, the \overline{QERR} output functions normally. The \overline{RESET} input has priority over the \overline{DCS} and \overline{CSR} control and, when driven low, forces the Qn and PPO outputs low and forces the \overline{QERR} output high. If the \overline{DCS} control functionality is not desired, the \overline{CSR} input can be hard-wired to ground, in which case the setup-time requirement for \overline{DCS} is the same as for the other D data inputs. To control the low-power mode with \overline{DCS} only, the \overline{CSR} input should be pulled up to V_{CC} through a pullup resistor.

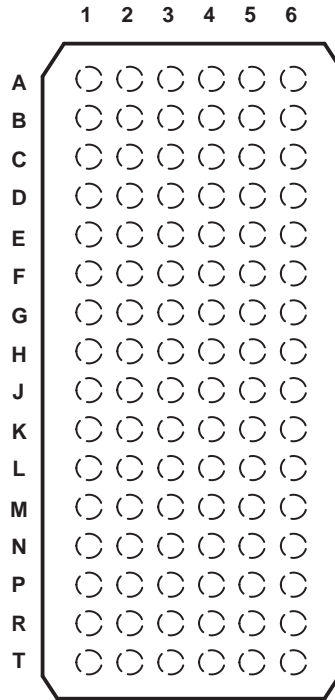
The two V_{REF} pins (A3 and T3) are connected together internally by approximately 150 Ω . However, it is necessary to connect only one of the two V_{REF} pins to the external V_{REF} power supply. An unused V_{REF} pin should be terminated with a V_{REF} coupling capacitor.



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**GKE PACKAGE
(TOP VIEW)**



terminal assignments for 1:1 register (C0 = 0, C1 = 0)

	1	2	3	4	5	6
A	D1 (DCKE)	PPO	V _{REF}	V _{CC}	Q1 (QCKE)	DNU
B	D2	D15	GND	GND	Q2	Q15
C	D3	D16	V _{CC}	V _{CC}	Q3	Q16
D	D4 (DODT)	\overline{QERR}	GND	GND	Q4 (QODT)	DNU
E	D5	D17	V _{CC}	V _{CC}	Q5	Q17
F	D6	D18	GND	GND	Q6	Q18
G	PAR_IN	\overline{RESET}	V _{CC}	V _{CC}	C1	C0
H	CLK	D7 (\overline{DCS})	GND	GND	Q7 (\overline{QCS})	DNU
J	\overline{CLK}	\overline{CSR}	V _{CC}	V _{CC}	NC	NC
K	D8	D19	GND	GND	Q8	Q19
L	D9	D20	V _{CC}	V _{CC}	Q9	Q20
M	D10	D21	GND	GND	Q10	Q21
N	D11	D22	V _{CC}	V _{CC}	Q11	Q22
P	D12	D23	GND	GND	Q12	Q23
R	D13	D24	V _{CC}	V _{CC}	Q13	Q24
T	D14	D25	V _{REF}	V _{CC}	Q14	Q25

Each pin name in parentheses indicates the DDR2 DIMM signal name.

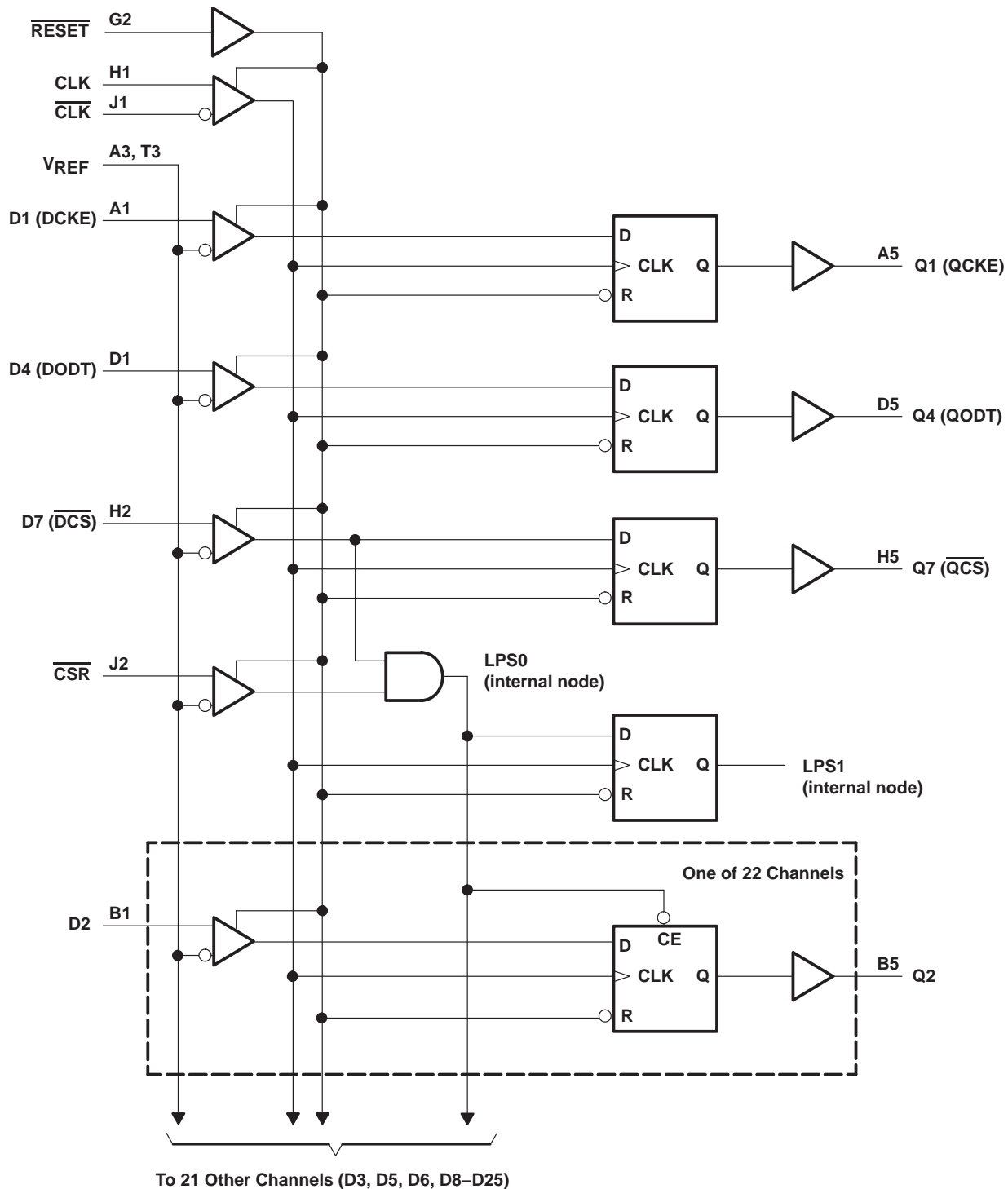
DNU – Do not use

NC – No internal connection

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logic diagram for 1:1 register configuration (positive logic); C0 = 0, C1 = 0

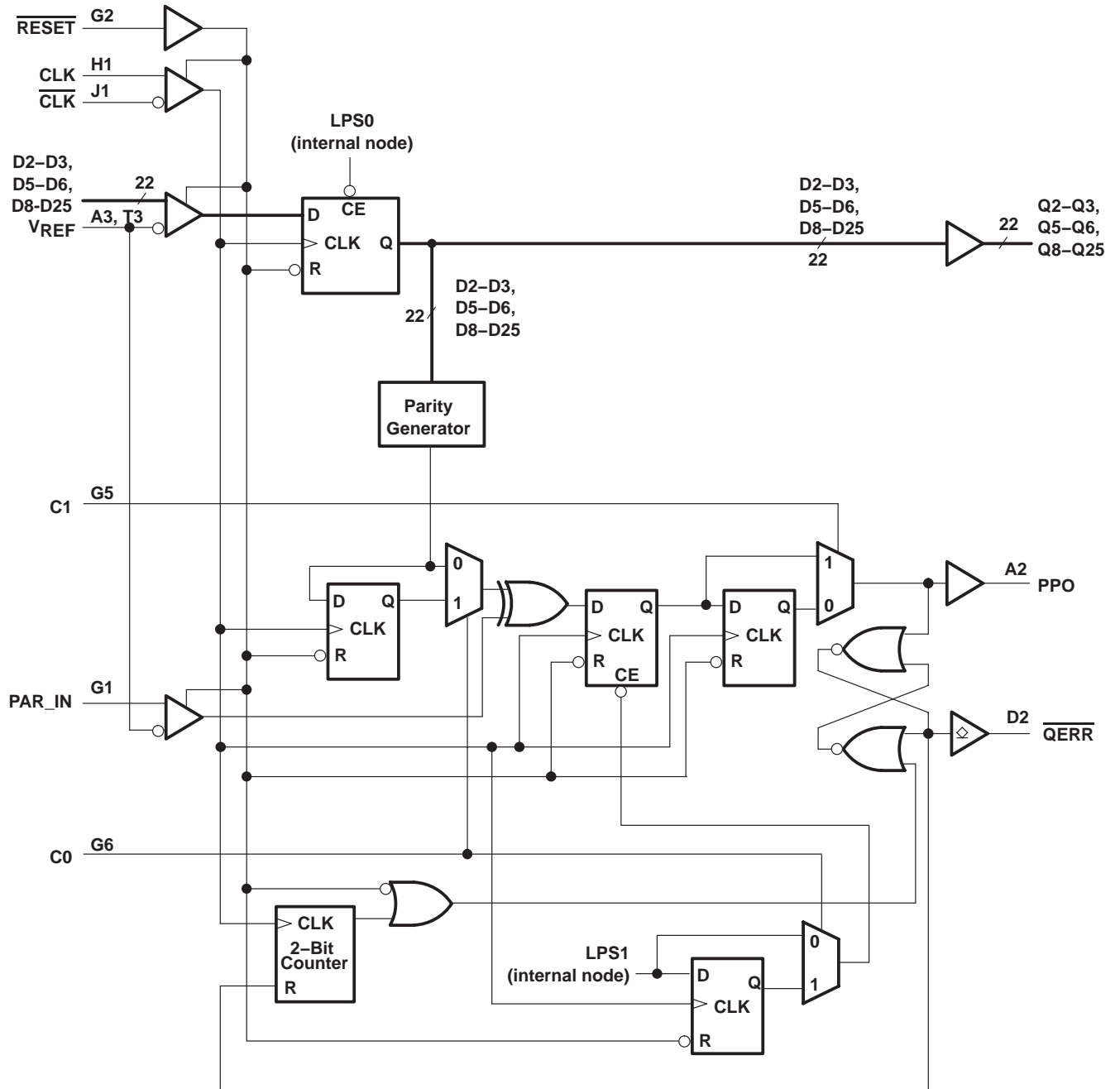


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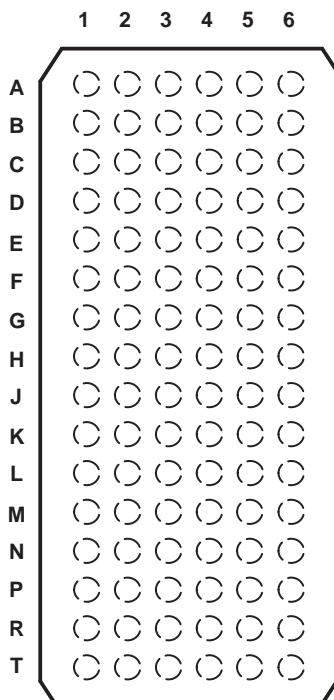
parity logic diagram for 1:1 register configuration (positive logic); C0 = 0, C1 = 0



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GKE PACKAGE
(TOP VIEW)



terminal assignments for 1:2 register-A (C0 = 0, C1 = 1)

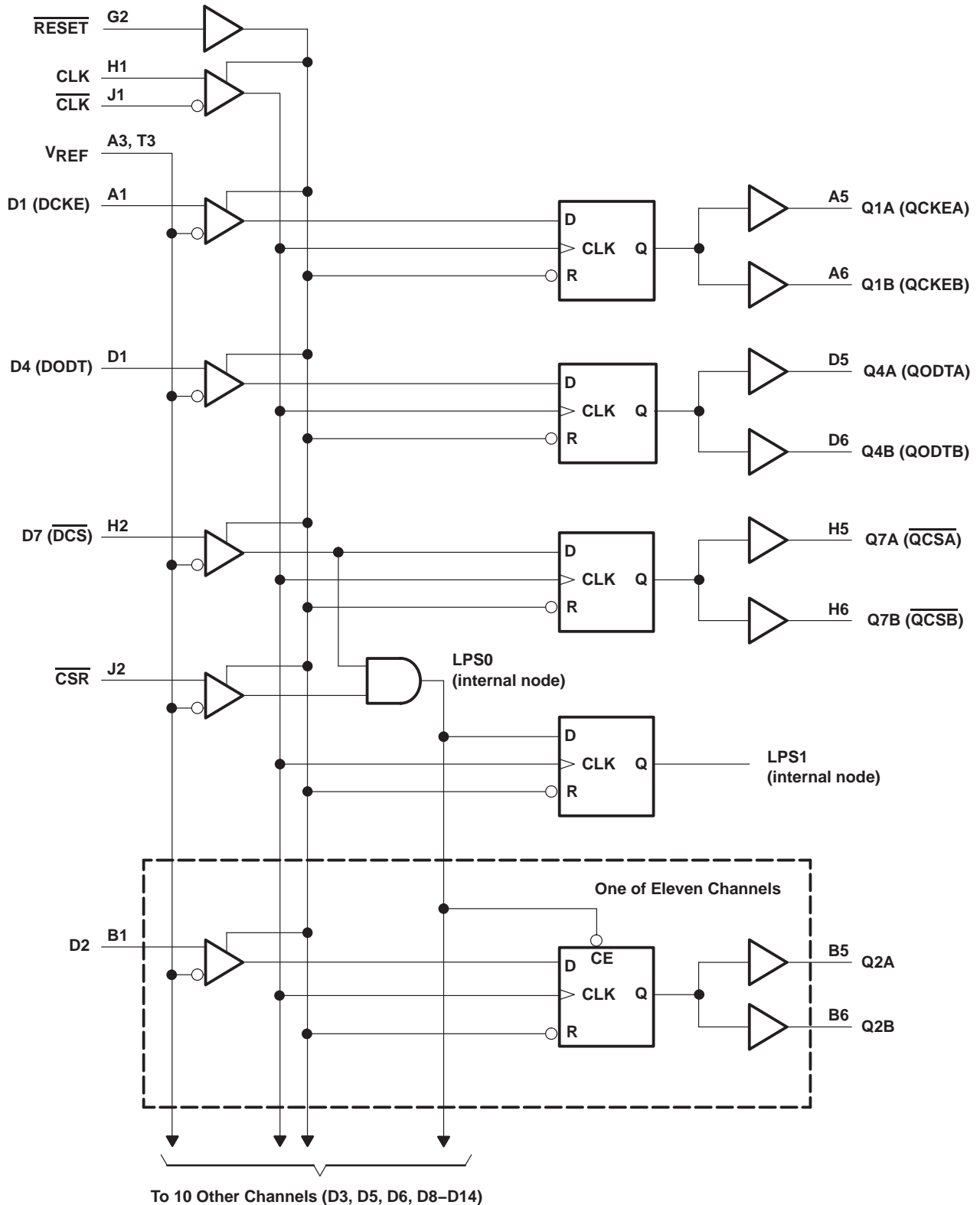
	1	2	3	4	5	6
A	D1 (DCKE)	PPO	VREF	VCC	Q1A (QCKEA)	Q1B (QCKEB)
B	D2	DNU	GND	GND	Q2A	Q2B
C	D3	DNU	VCC	VCC	Q3A	Q3B
D	D4 (DODT)	\overline{QERR}	GND	GND	Q4A (QODTA)	Q4B (QODTB)
E	D5	DNU	VCC	VCC	Q5A	Q5B
F	D6	DNU	GND	GND	Q6A	Q6B
G	PAR_IN	\overline{RESET}	VCC	VCC	C1	C0
H	CLK	D7 (\overline{DCS})	GND	GND	Q7A (QCSA)	Q7B (QCSB)
J	\overline{CLK}	\overline{CSR}	VCC	VCC	NC	NC
K	D8	DNU	GND	GND	Q8A	Q8B
L	D9	DNU	VCC	VCC	Q9A	Q9B
M	D10	DNU	GND	GND	Q10A	Q10B
N	D11	DNU	VCC	VCC	Q11A	Q11B
P	D12	DNU	GND	GND	Q12A	Q12B
R	D13	DNU	VCC	VCC	Q13A	Q13B
T	D14	DNU	VREF	VCC	Q14A	Q14B

Each pin name in parentheses indicates the DDR2 DIMM signal name.
DNU – Do not use
NC – No internal connection

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logic diagram for 1:2 register-A configuration (positive logic); C0 = 0, C1 = 1



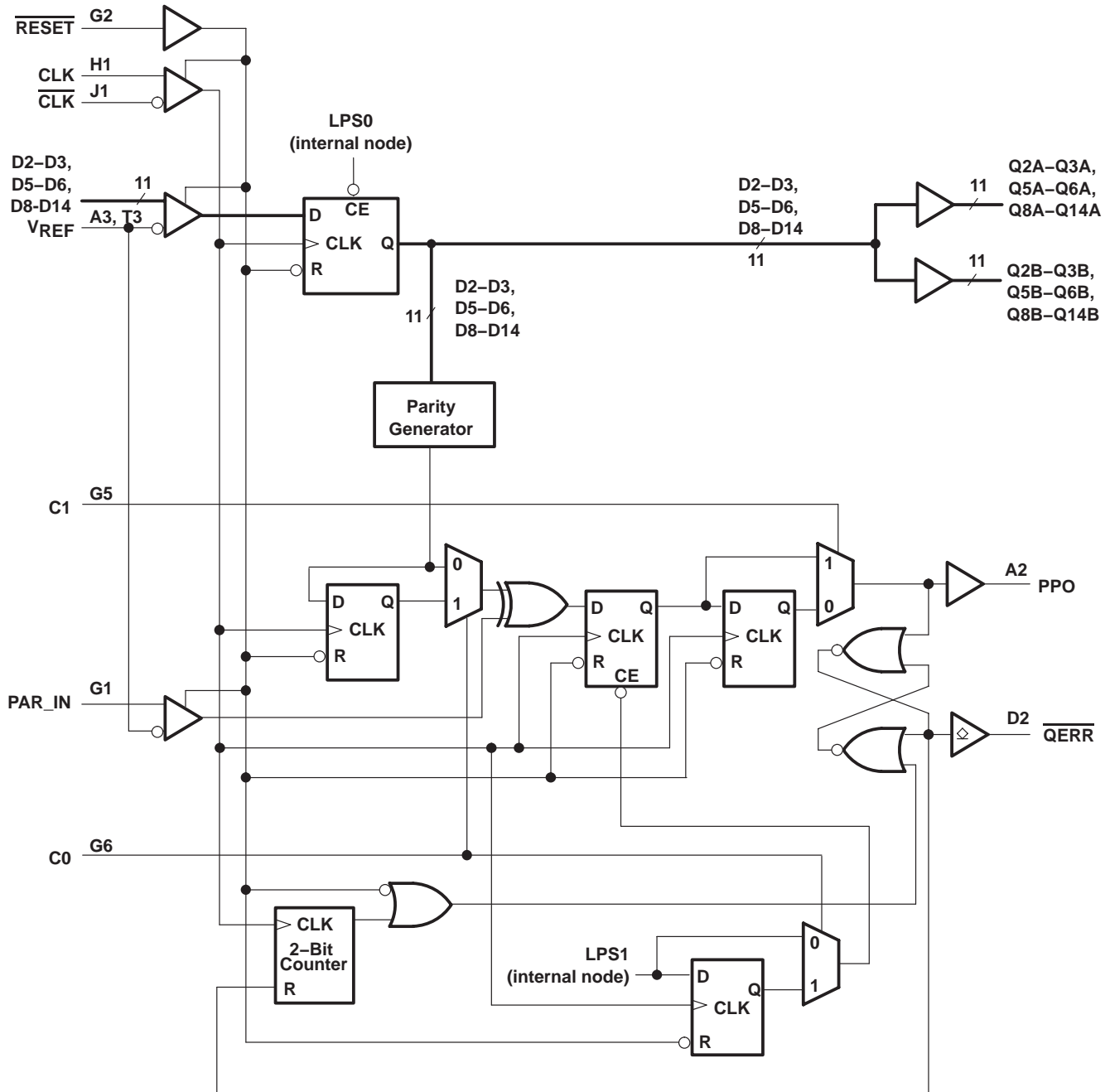
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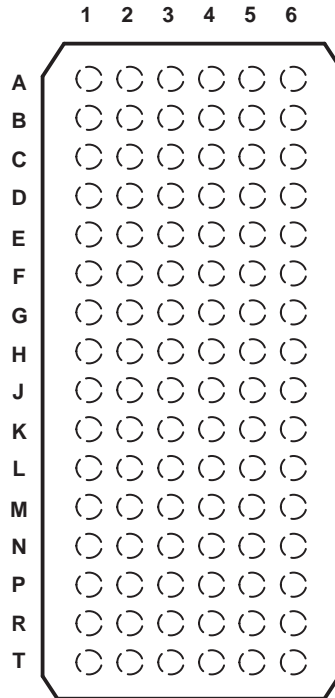
parity logic diagram for 1:2 register-A configuration (positive logic); C0 = 0, C1 = 1



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GKE PACKAGE
(TOP VIEW)



terminal assignments for 1:2 register-B (C0 = 1, C1 = 1)

	1	2	3	4	5	6
A	D1	PPO	V _{REF}	V _{CC}	Q1A	Q1B
B	D2	DNU	GND	GND	Q2A	Q2B
C	D3	DNU	V _{CC}	V _{CC}	Q3A	Q3B
D	D4	\overline{QERR}	GND	GND	Q4A	Q4B
E	D5	DNU	V _{CC}	V _{CC}	Q5A	Q5B
F	D6	DNU	GND	GND	Q6A	Q6B
G	PAR_IN	\overline{RESET}	V _{CC}	V _{CC}	C1	C0
H	CLK	D7 (\overline{DCS})	GND	GND	$\overline{Q7A}$ (QCSA)	$\overline{Q7B}$ (QCSB)
J	\overline{CLK}	\overline{CSR}	V _{CC}	V _{CC}	NC	NC
K	D8	DNU	GND	GND	Q8A	Q8B
L	D9	DNU	V _{CC}	V _{CC}	Q9A	Q9B
M	D10	DNU	GND	GND	Q10A	Q10B
N	D11 (DODT)	DNU	V _{CC}	V _{CC}	Q11A (QODTA)	Q11B (QODTB)
P	D12	DNU	GND	GND	Q12A	Q12B
R	D13	DNU	V _{CC}	V _{CC}	Q13A	Q13B
T	D14 (DCKE)	DNU	V _{REF}	V _{CC}	Q14A (QCKEA)	Q14B (QCKEB)

Each pin name in parentheses indicates the DDR2 DIMM signal name.

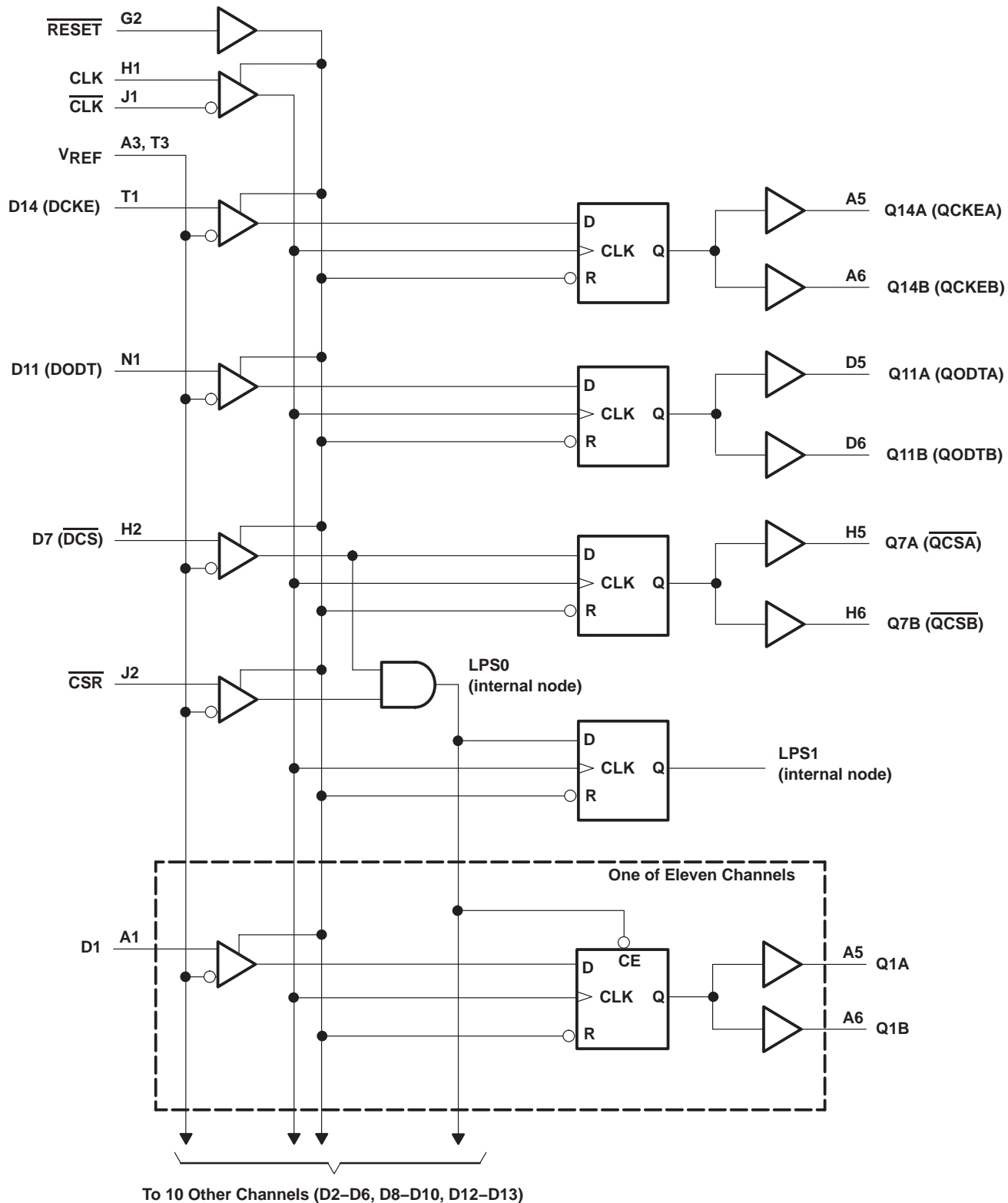
DNU – Do not use

NC – No internal connection

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logic diagram for 1:2 register-B configuration (positive logic); C0 = 1, C1 = 1

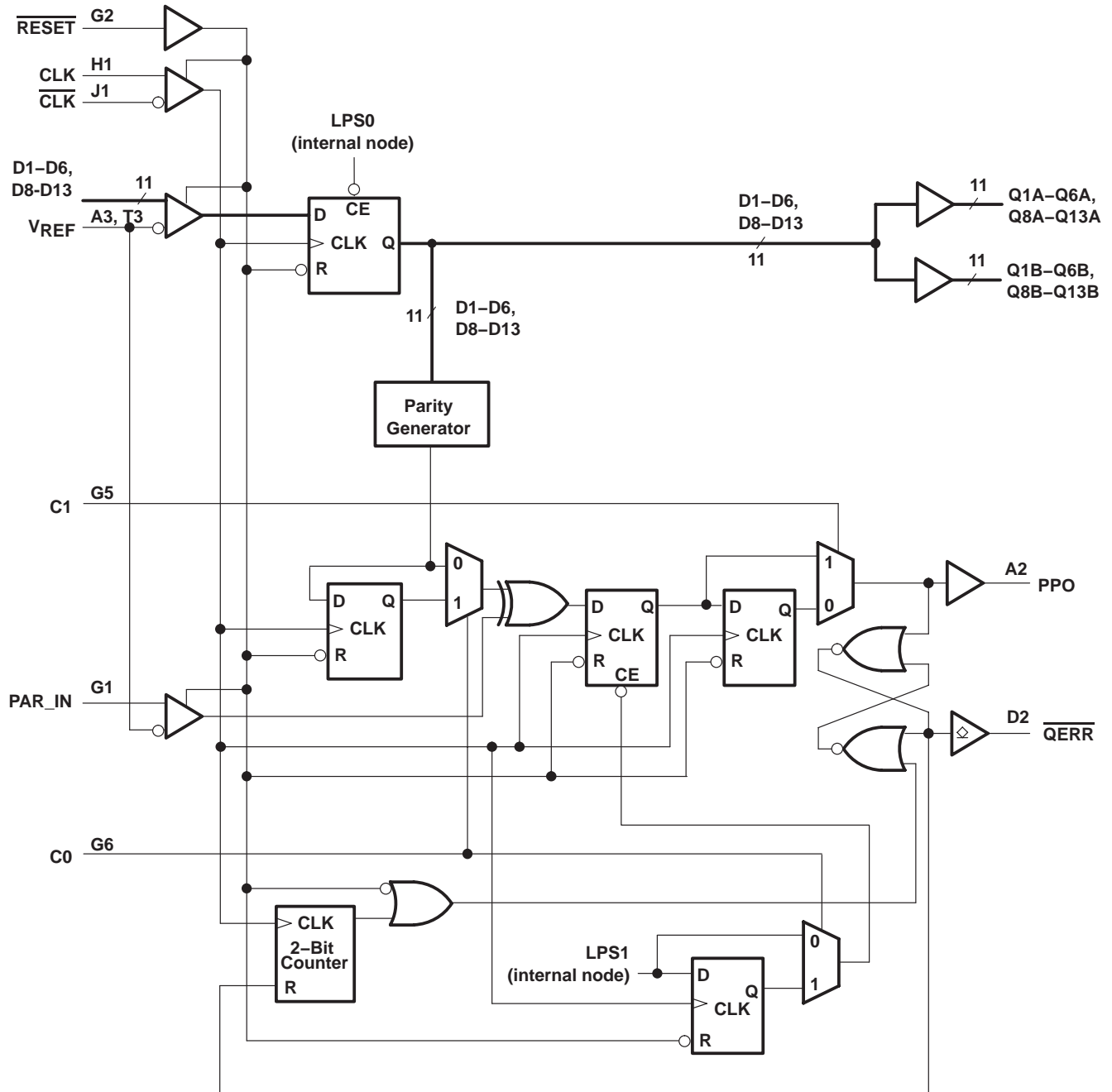


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parity logic diagram for 1:2 register-B configuration (positive logic); C0 = 1, C1 = 1



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TERMINAL FUNCTIONS

TERMINAL NAME	DESCRIPTION	ELECTRICAL CHARACTERISTICS
GND	Ground	Ground input
V _{CC}	Power-supply voltage	1.8 V nominal
V _{REF}	Input reference voltage	0.9 V nominal
CLK	Positive master clock input	Differential input
$\overline{\text{CLK}}$	Negative master clock input	Differential input
C0, C1	Configuration control input. Register A or Register B and 1:1 mode or 1:2 mode select.	LVC MOS input
$\overline{\text{RESET}}$	Asynchronous reset input. Resets registers and disables V _{REF} , data, and clock differential-input receivers. When $\overline{\text{RESET}}$ is low, all Q outputs are forced low and the $\overline{\text{QERR}}$ output is forced high.	LVC MOS input
D1–D25	Data input. Clocked in on the crossing of the rising edge of CLK and the falling edge of $\overline{\text{CLK}}$.	SSTL ₁₈ inputs
$\overline{\text{CSR}}$, $\overline{\text{DCS}}$	Chip select inputs. Disables D1–D25 [†] outputs switching when both inputs are high	SSTL ₁₈ inputs
DOT	The outputs of this register bit will not be suspended by the $\overline{\text{DCS}}$ and $\overline{\text{CSR}}$ control.	SSTL ₁₈ input
DCKE	The outputs of this register bit will not be suspended by the $\overline{\text{DCS}}$ and $\overline{\text{CSR}}$ control.	SSTL ₁₈ input
PAR_IN	Parity input. Arrives one clock cycle after the corresponding data input.	SSTL ₁₈ input
Q1–Q25 [‡]	Data outputs that are suspended by the $\overline{\text{DCS}}$ and $\overline{\text{CSR}}$ control.	1.8 V CMOS outputs
PPO	Partial parity out. Indicates odd parity of inputs D1–D25. [†]	1.8 V CMOS output
$\overline{\text{QCS}}$	Data output that will not be suspended by the $\overline{\text{DCS}}$ and $\overline{\text{CSR}}$ control	1.8 V CMOS output
QODT	Data output that will not be suspended by the $\overline{\text{DCS}}$ and $\overline{\text{CSR}}$ control	1.8 V CMOS output
QCKE	Data output that will not be suspended by the $\overline{\text{DCS}}$ and $\overline{\text{CSR}}$ control	1.8 V CMOS output
$\overline{\text{QERR}}$	Output error bit. Timing is determined by the device mode.	Open-drain output
NC	No internal connection	
DNU	Do not use. Inputs are in standby-equivalent mode, and outputs are driven low.	

[†] Data inputs = D2, D3, D5, D6, D8–D25 when C0 = 0 and C1 = 0
 Data inputs = D2, D3, D5, D6, D8–D14 when C0 = 0 and C1 = 1
 Data inputs = D1–D6, D8–D10, D12, D13 when C0 = 1 and C1 = 1.

[‡] Data outputs = Q2, Q3, Q5, Q6, Q8–Q25 when C0 = 0 and C1 = 0
 Data outputs = Q2, Q3, Q5, Q6, Q8–Q14 when C0 = 0 and C1 = 1
 Data outputs = Q1–Q6, Q8–Q10, Q12, Q13 when C0 = 1 and C1 = 1.

FUNCTION TABLES

INPUTS						OUTPUTS
$\overline{\text{RESET}}$	$\overline{\text{DCS}}$	$\overline{\text{CSR}}$	CLK	$\overline{\text{CLK}}$	D _n	Q _n
H	L	X	↑	↓	L	L
H	L	X	↑	↓	H	H
H	X	L	↑	↓	L	L
H	X	L	↑	↓	H	H
H	H	H	↑	↓	X	Q ₀
H	X	X	L or H	L or H	X	Q ₀
L	X or floating	X or floating	X or floating	X or floating	X or floating	L



Function Tables (Continued)

INPUTS				OUTPUTS
$\overline{\text{RESET}}$	CLK	$\overline{\text{CLK}}$	$\overline{\text{DCS}}$, $\overline{\text{DCS}}$, DODT	$\overline{\text{QCKE}}$, $\overline{\text{QCS}}$, QODT
H	↑	↓	H	H
H	↑	↓	L	L
H	L or H	L or H	X	Q ₀
L	X or floating	X or floating	X or floating	L

PARITY AND STANDBY FUNCTION

INPUTS							OUTPUTS	
$\overline{\text{RESET}}$	CLK	$\overline{\text{CLK}}$	$\overline{\text{DCS}}$	$\overline{\text{CSR}}$	Σ OF INPUTS = H D1–D25 [†]	PAR_IN [‡]	PPO	$\overline{\text{QERR}}^{\S}$
H	↑	↓	L	X	Even	L	L	H
H	↑	↓	L	X	Odd	L	H	L
H	↑	↓	L	X	Even	H	H	L
H	↑	↓	L	X	Odd	H	L	H
H	↑	↓	H	L	Even	L	L	H
H	↑	↓	H	L	Odd	L	H	L
H	↑	↓	H	L	Even	H	H	L
H	↑	↓	H	L	Odd	H	L	H
H	↑	↓	H	H	X	X	PPO ₀	$\overline{\text{QERR}}_0$
H	L or H	L or H	X	X	X	X	PPO ₀	$\overline{\text{QERR}}_0$
L	X or floating	X or floating	X or floating	X or floating	X	X or floating	L	H

[†] Data inputs = D2–D3, D5–D6, D8–D25 when C0 = 0 and C1 = 0

Data inputs = D2–D3, D5–D6, D8–D14 when C0 = 0 and C1 = 1

Data inputs = D1–D6, D8–D13 when C0 = 1 and C1 = 1

[‡] PAR_IN arrives one clock cycle (C0 = 0) or two clock cycles (C1 = 1) after the data to which it applies.

[§] This transition assumes that QERR is high at the crossing of CLK going high and CLK going low. If QERR goes low, it stays latched low for a minimum of two clock cycles or until RESET is driven low. If two or more consecutive parity errors occur, the QERR output is driven low and latched low for a clock duration equal to the parity duration or until RESET is driven low.

PARITY ERROR DETECT IN LOW-POWER MODE[¶]

ERROR POSITION	1:1 MODE (C0 = 0, C1 = 0)		1:2 REGISTER-A MODE (C0 = 0, C1 = 1)		1:2 REGISTER-B MODE (C0 = 0, C1 = 0)		CASCADED MODE (Registers A and B)	
	PPO DURATION	$\overline{\text{QERR}}$ DURATION	PPO DURATION	$\overline{\text{QERR}}$ DURATION	PPO DURATION	$\overline{\text{QERR}}$ DURATION	PPO DURATION	$\overline{\text{QERR}}$ DURATION
n – 4	1 Cycle	2 Cycles	1 Cycle	2 Cycles	1 Cycle	2 Cycles	1 Cycle	2 Cycles
n – 3	1 Cycle	2 Cycles	1 Cycle	2 Cycles	1 Cycle	2 Cycles	1 Cycle	2 Cycles
n – 2	1 Cycle	2 Cycles	1 Cycle	2 Cycles	1 Cycle	2 Cycles	1 Cycle	2 Cycles
n – 1	LPM + 2	LPM + 2	LPM + 1	LPM + 1	LPM + 2	LPM + 2	LPM + 2	LPM + 2
n	Not detected	Not detected	Not detected	Not detected	Not detected	Not detected	Not detected	Not detected

[¶] If a parity error occurs before the device enters the low-power mode (LPM), the behavior of PPO and $\overline{\text{QERR}}$ is dependent on the mode of the device and the position of the parity error occurrence. This table illustrates the low-power-mode effect on parity detect. The low-power mode is activated on the n clock cycle when $\overline{\text{DCS}}$ and $\overline{\text{CSR}}$ go high.

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absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage range, V_{CC}	-0.5 V to 2.5 V
Input voltage range, V_I (see Notes 1 and 2)	-0.5 V to 2.5 V
Output voltage range, V_O (see Notes 1 and 2)	-0.5 V to $V_{CC} + 0.5$ V
Input clamp current, I_{IK} ($V_I < 0$ or $V_I > V_{CC}$)	± 50 mA
Output clamp current, I_{OK} ($V_O < 0$ or $V_O > V_{CC}$)	± 50 mA
Continuous output current, I_O ($V_O = 0$ to V_{CC})	± 50 mA
Continuous current through each V_{CC} or GND	± 100 mA
Package thermal impedance, θ_{JA} (see Note 3)	36°C/W
Storage temperature range, T_{stg}	-65°C to 150°C

† Stresses beyond those listed under “absolute maximum ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under “recommended operating conditions” is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. The input and output negative-voltage ratings may be exceeded if the input and output clamp-current ratings are observed.
 2. This value is limited to 2.5 V maximum.
 3. The package thermal impedance is calculated in accordance with JESD 51-7.

recommended operating conditions (see Note 4)

		MIN	NOM	MAX	UNIT
V_{CC}	Supply voltage	1.7		1.9	V
V_{REF}	Reference voltage	$0.49 \times V_{CC}$	$0.5 \times V_{CC}$	$0.51 \times V_{CC}$	V
V_{TT}	Termination voltage	$V_{REF} - 40$ mV	V_{REF}	$V_{REF} + 40$ mV	V
V_I	Input voltage	0		V_{CC}	V
V_{IH}	AC high-level input voltage	Data inputs, \overline{CSR} , $\overline{PAR_IN}$	$V_{REF} + 250$ mV		V
V_{IL}	AC low-level input voltage	Data inputs, \overline{CSR} , $\overline{PAR_IN}$		$V_{REF} - 250$ mV	V
V_{IH}	DC high-level input voltage	Data inputs, \overline{CSR} , $\overline{PAR_IN}$	$V_{REF} + 125$ mV		V
V_{IL}	DC low-level input voltage	Data inputs, \overline{CSR} , $\overline{PAR_IN}$		$V_{REF} - 125$ mV	V
V_{IH}	High-level input voltage	\overline{RESET} , C_n	$0.65 \times V_{CC}$		V
V_{IL}	Low-level input voltage	\overline{RESET} , C_n		$0.35 \times V_{CC}$	V
V_{ICR}	Common-mode input voltage range	\overline{CLK} , \overline{CLK}	0.675	1.125	V
$V_{I(PP)}$	Peak-to-peak input voltage	\overline{CLK} , \overline{CLK}	600		mV
I_{OH}	High-level output current	Q outputs, PPO		-8	mA
I_{OL}	Low-level output current	Q outputs, PPO		8	mA
		\overline{QERR} output		8	
T_A	Operating free-air temperature	0		70	°C

NOTE 4: The \overline{RESET} and C_n inputs of the device must be held at valid logic voltage levels (not floating) to ensure proper device operation. The differential inputs must not be floating unless \overline{RESET} is low. Refer to the TI application report, *Implications of Slow or Floating CMOS Inputs*, literature number SCBA004.



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electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	V _{CC}	MIN	TYP†	MAX	UNIT
V _{OH}	Q outputs, PPO	I _{OH} = -100 μA	1.7 V to 1.9 V	V _{CC} -0.2			V
		I _{OH} = -6 mA	1.7 V	1.3			
V _{OL}	Q outputs, PPO	I _{OL} = 100 μA	1.7 V to 1.9 V	0.2			V
		I _{OL} = 6 mA	1.7 V	0.4			
	\overline{QERR} output	I _{OL} = 8 mA	1.7 V	0.35			
I _I	All inputs‡	V _I = V _{CC} or GND	1.9 V	±5			μA
I _{OZ}	\overline{QERR} output	V _O = V _{CC} or GND	1.9 V	±10			μA
I _{CC}	Static standby	$\overline{RESET} = GND$	1.9 V	100			μA
	Static operating	$\overline{RESET} = V_{CC}, V_I = V_{IH(AC)} \text{ or } V_{IL(AC)}$		50			mA
I _{CCD}	Dynamic operating – clock only	$\overline{RESET} = V_{CC}, V_I = V_{IH(AC)} \text{ or } V_{IL(AC)},$ CLK and CLK switching 50% duty cycle	1.8 V	42			μA/ MHz
	Dynamic operating – per each data input, 1:1 configuration	$\overline{RESET} = V_{CC}, V_I = V_{IH(AC)} \text{ or } V_{IL(AC)},$ CLK and CLK switching 50% duty cycle,		18			μA/ clock MHz/ D input
	Dynamic operating – per each data input, 1:2 configuration	one data input switching at one-half clock frequency, 50% duty cycle		35			
I _{CCDLP}	Chip-select-enabled low-power active mode – clock only	$\overline{RESET} = V_{CC}, V_I = V_{IH(AC)} \text{ or } V_{IL(AC)},$ CLK and CLK switching 50% duty cycle	1.8 V	42			μA/ MHz
	Chip-select-enabled low-power active mode – 1:1 configuration	$\overline{RESET} = V_{CC}, V_I = V_{IH(AC)} \text{ or } V_{IL(AC)},$ CLK and \overline{CLK} switching 50% duty cycle,		2			μA/ clock MHz/ D input
	Chip-select-enabled low-power active mode – 1:2 configuration	one data input switching at one-half clock frequency, 50% duty cycle		2			
C _i	Data inputs, \overline{CSR} , PAR_IN	V _I = V _{REF} ± 250 mV	1.8 V	2.5	3	3.5	pF
	CLK, \overline{CLK}	V _{ICR} = 0.9 V, V _{I(PP)} = 600 mV		2	3		
	\overline{RESET}	V _I = V _{CC} or GND		2.5			

† All typical values are at V_{CC} = 1.8 V, T_A = 25°C.

‡ Each V_{REF} pin (A3 or T3) should be tested independently, with the other (untested) pin open.



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timing requirements over recommended operating free-air temperature range (unless otherwise noted) (see Figure 1 and Note 5)

		$V_{CC} = 1.8\text{ V} \pm 0.1\text{ V}$		UNIT
		MIN	MAX	
f_{clock}	Clock frequency	500		MHz
t_w	Pulse duration, CLK, $\overline{\text{CLK}}$ high or low	1		ns
t_{act}	Differential inputs active time (see Note 6)	10		ns
t_{inact}	Differential inputs inactive time (see Note 7)	15		ns
t_{su}	Setup time	$\overline{\text{DCS}}$ before CLK \uparrow , $\overline{\text{CLK}}\downarrow$, $\overline{\text{CSR}}$ high; $\overline{\text{CSR}}$ before CLK \uparrow , $\overline{\text{CLK}}\downarrow$, $\overline{\text{DCS}}$ high		ns
		$\overline{\text{DCS}}$ before CLK \uparrow , $\overline{\text{CLK}}\downarrow$, $\overline{\text{CSR}}$ low		
		DODT, DCKE, and Data before CLK \uparrow , $\overline{\text{CLK}}\downarrow$		
		PAR_IN before CLK \uparrow , $\overline{\text{CLK}}\downarrow$		
t_h	Hold time	$\overline{\text{DCS}}$, DODT, DCKE, and Data after CLK \uparrow , $\overline{\text{CLK}}\downarrow$		ns
		PAR_IN after CLK \uparrow , $\overline{\text{CLK}}\downarrow$		

NOTES: 5. All inputs slew rate is 1 V/ns \pm 20%.

6. V_{REF} must be held at a valid input level, and data inputs must be held low for a minimum time of t_{act} max, after $\overline{\text{RESET}}$ is taken high.

7. V_{REF} , data, and clock inputs must be held at valid voltage levels (not floating) for a minimum time of t_{inact} max, after $\overline{\text{RESET}}$ is taken low.

switching characteristics over recommended operating free-air temperature range (unless otherwise noted)

PARAMETER		FROM (INPUT)	TO (OUTPUT)	$V_{CC} = 1.8\text{ V} \pm 0.1\text{ V}$		UNIT
				MIN	MAX	
f_{max} (see Figure 1)				500		MHz
t_{pdm}^\dagger (see Figure 1)		CLK and $\overline{\text{CLK}}$	Q	1.4	2.5	ns
t_{pd}	see Figure 4	CLK and $\overline{\text{CLK}}$	PPO	0.6	1.6	ns
t_{PLH}	see Figure 3	CLK and $\overline{\text{CLK}}$	$\overline{\text{QERR}}$	1.2	3	ns
t_{PHL}				1	2.4	
t_{pdmss}^\dagger (see Figure 1)		CLK and $\overline{\text{CLK}}$	Q	2.7		ns
t_{RPHL}^\dagger (see Figure 1)		$\overline{\text{RESET}}$	Q	3		ns
t_{RPHL} (see Figure 4)			PPO	3		
t_{RPLH} (see Figure 3)			$\overline{\text{QERR}}$	3		

† Includes 350-ps test-load transmission-line delay.

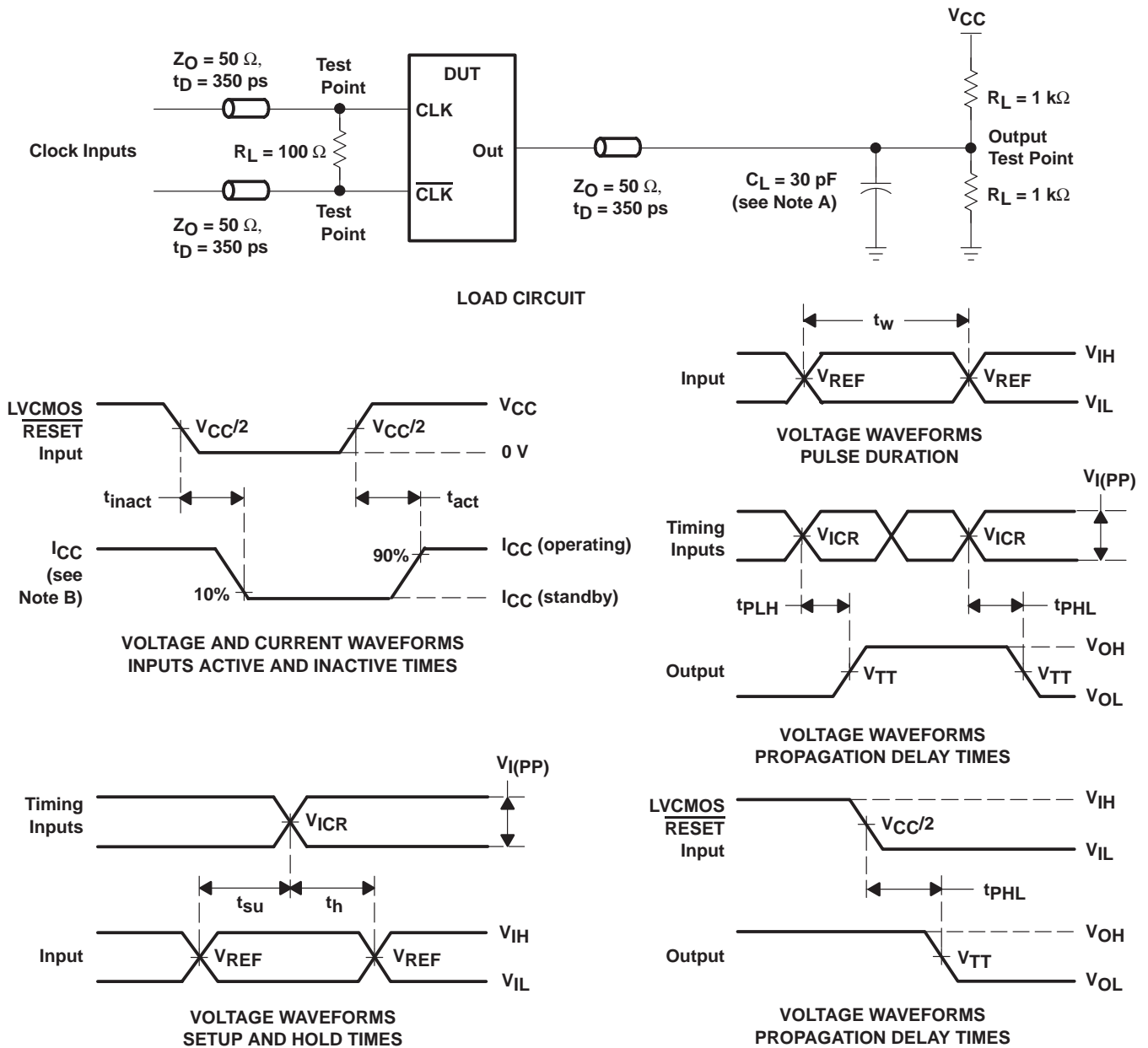
output slew rates over recommended operating free-air temperature range (unless otherwise noted) (see Figure 2)

PARAMETER	FROM	TO	$V_{CC} = 1.8\text{ V} \pm 0.1\text{ V}$		UNIT
			MIN	MAX	
dV/dt_r	20%	80%	1.9	4.9	V/ns
dV/dt_f	80%	20%	1.9	4.9	V/ns
dV/dt_{Δ}^\ddagger	20% or 80%	80% or 20%	1.5		V/ns

‡ Difference between dV/dt_r (rising edge rate) and dV/dt_f (falling edge rate).



PARAMETER MEASUREMENT INFORMATION



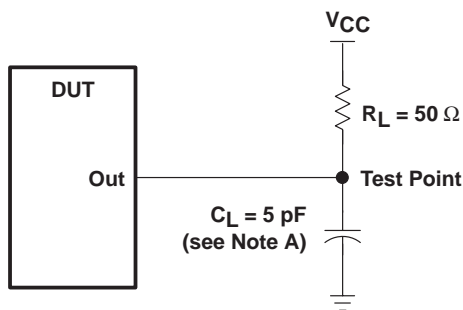
- NOTES:
- A. C_L includes probe and jig capacitance.
 - B. I_{CC} tested with clock and data inputs held at V_{CC} or GND, and $I_O = 0$ mA.
 - C. All input pulses are supplied by generators having the following characteristics: PRR ≤ 10 MHz, $Z_O = 50 \Omega$, input slew rate = 1 V/ns $\pm 20\%$ (unless otherwise noted).
 - D. The outputs are measured one at a time, with one transition per measurement.
 - E. $V_{REF} = V_{TT} = V_{CC}/2$
 - F. $V_{IH} = V_{REF} + 250$ mV (ac voltage levels) for differential inputs. $V_{IH} = V_{CC}$ for LVC MOS input.
 - G. $V_{IL} = V_{REF} - 250$ mV (ac voltage levels) for differential inputs. $V_{IL} = \text{GND}$ for LVC MOS input.
 - H. $V_{I(PP)} = 600$ mV
 - I. t_{PLH} and t_{PHL} are the same as t_{pD} .

Figure 1. Data Output Load Circuit and Voltage Waveforms

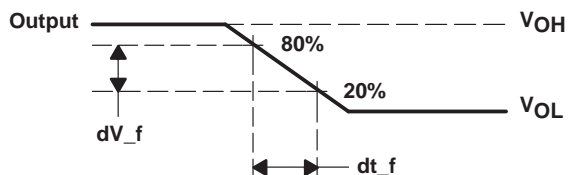
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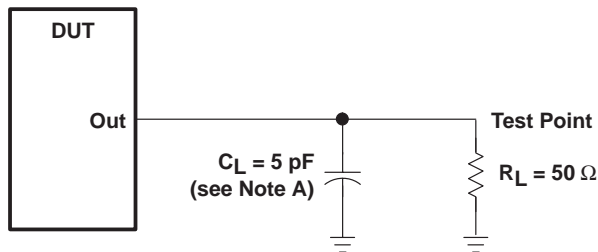
PARAMETER MEASUREMENT INFORMATION



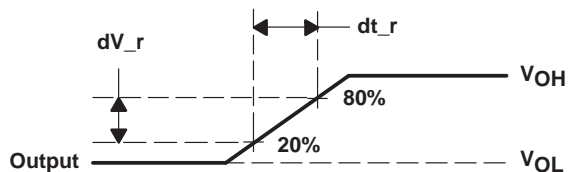
LOAD CIRCUIT
HIGH-TO-LOW SLEW-RATE MEASUREMENT



VOLTAGE WAVEFORMS
HIGH-TO-LOW SLEW-RATE MEASUREMENT



LOAD CIRCUIT
LOW-TO-HIGH SLEW-RATE MEASUREMENT

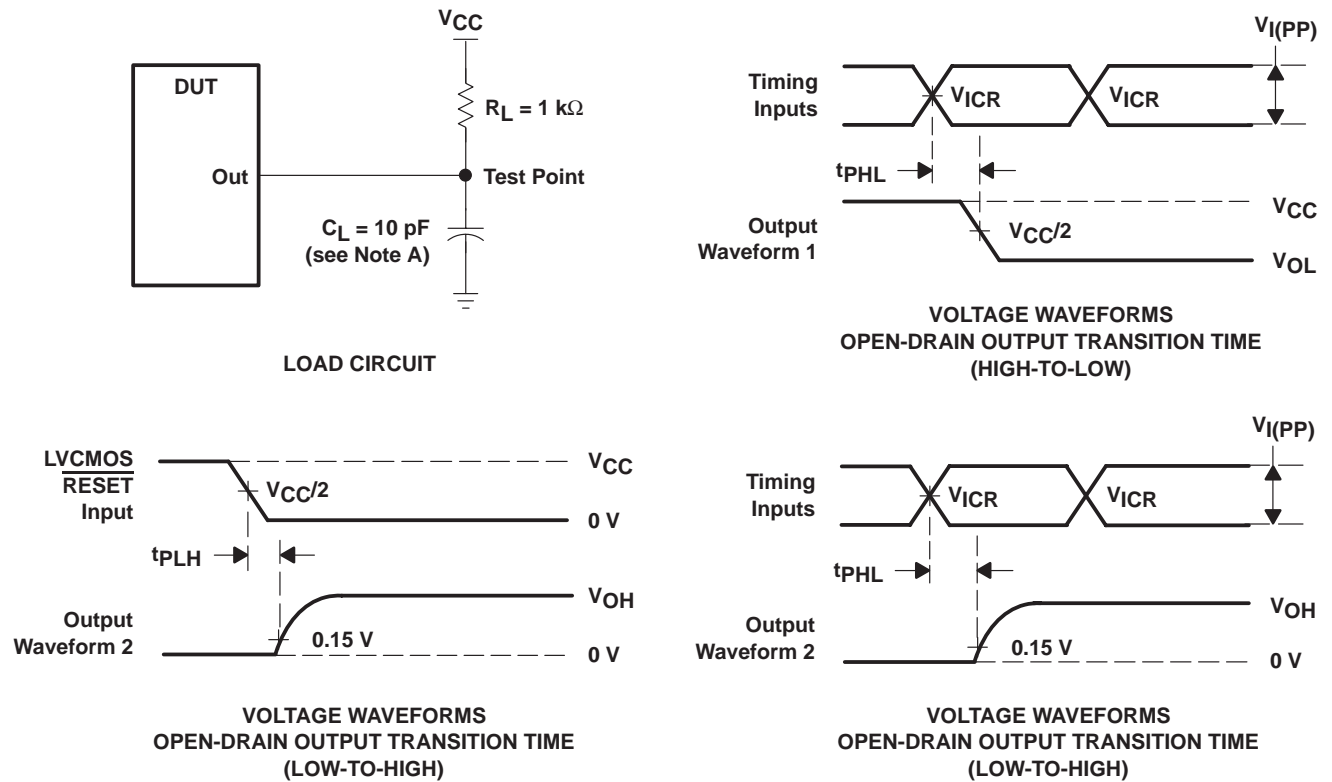


VOLTAGE WAVEFORMS
LOW-TO-HIGH SLEW-RATE MEASUREMENT

- NOTES: A. C_L includes probe and jig capacitance.
 B. All input pulses are supplied by generators having the following characteristics: $PRR \leq 10$ MHz, $Z_O = 50 \Omega$, input slew rate = 1 V/ns $\pm 20\%$ (unless otherwise specified).

Figure 2. Data Output Slew-Rate Measurement Information

PARAMETER MEASUREMENT INFORMATION



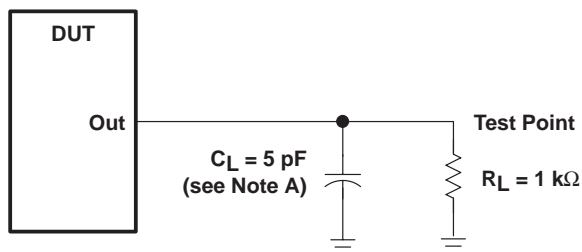
- NOTES:
- A. C_L includes probe and jig capacitance.
 - B. All input pulses are supplied by generators having the following characteristics: $PRR \leq 10\text{ MHz}$, $Z_O = 50\ \Omega$, input slew rate = $1\text{ V/ns} \pm 20\%$ (unless otherwise noted).
 - C. t_{PLH} and t_{PHL} are the same as t_{pd} .

Figure 3. Error Output Load Circuit and Voltage Waveforms

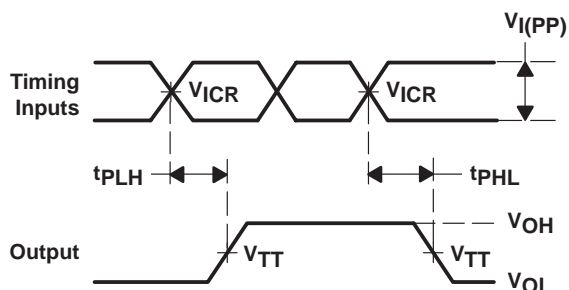
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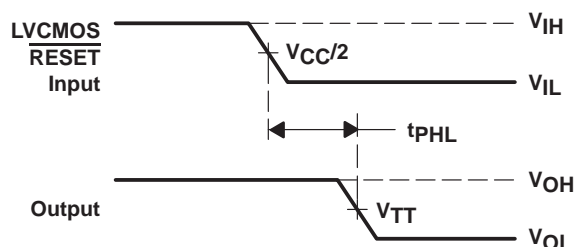
PARAMETER MEASUREMENT INFORMATION



LOAD CIRCUIT



VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES



VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES

- NOTES: A. C_L includes probe and jig capacitance.
 B. All input pulses are supplied by generators having the following characteristics: $PRR \leq 10 \text{ MHz}$, $Z_O = 50 \Omega$, input slew rate = $1 \text{ V/ns} \pm 20\%$ (unless otherwise noted).
 C. $V_{REF} = V_{TT} = V_{CC}/2$
 D. $V_{IH} = V_{REF} + 250 \text{ mV}$ (ac voltage levels) for differential inputs. $V_{IH} = V_{CC}$ for LVCMOS input.
 E. $V_{IL} = V_{REF} - 250 \text{ mV}$ (ac voltage levels) for differential inputs. $V_{IL} = \text{GND}$ for LVCMOS input.
 F. $V_{I(PP)} = 600 \text{ mV}$
 G. t_{PLH} and t_{PHL} are the same as t_{pd} .

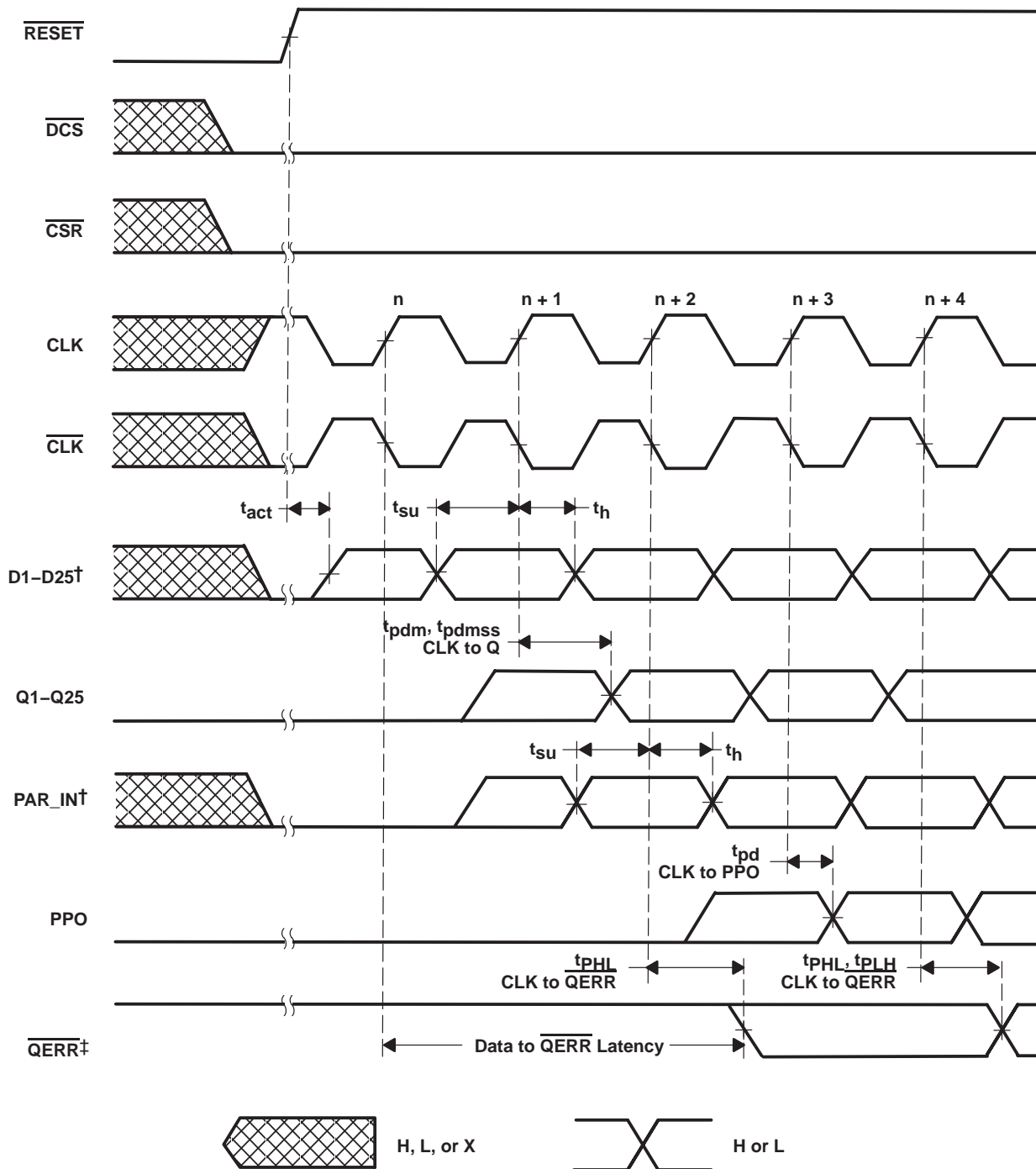
Figure 4. Partial-Parity-Out Load Circuit and Voltage Waveforms

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timing diagram for SN74SSTU32866 used as a single device; C0 = 0, C1 = 0
(RESET switches from L to H)



† After RESET is switched from low to high, all data and PAR_IN input signals must be set and held low for a minimum time of t_{act} max, to avoid false error.

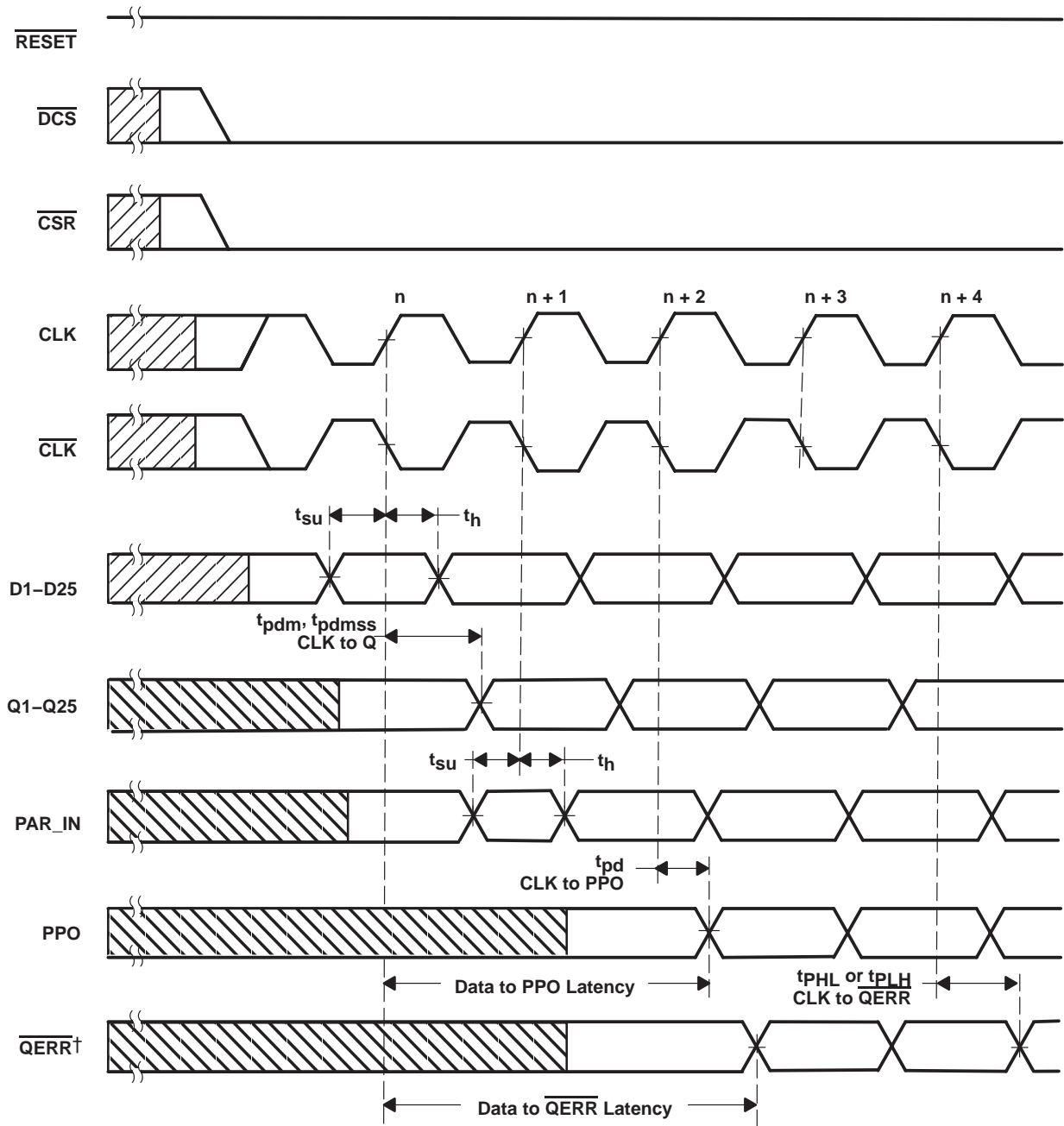
‡ If the data is clocked in on the n clock pulse, the QERR output signal will be generated on the n + 2 clock pulse, and it will be valid on the n + 3 clock pulse.

SN74SSTU32866

25-BIT CONFIGURABLE REGISTERED BUFFER WITH ADDRESS-PARITY TEST

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timing diagram for SN74SSTU32866 used as a single device; C0 = 0, C1 = 0
(RESET = H)



Unknown input event



Output signal is dependent on the prior unknown input event



H or L

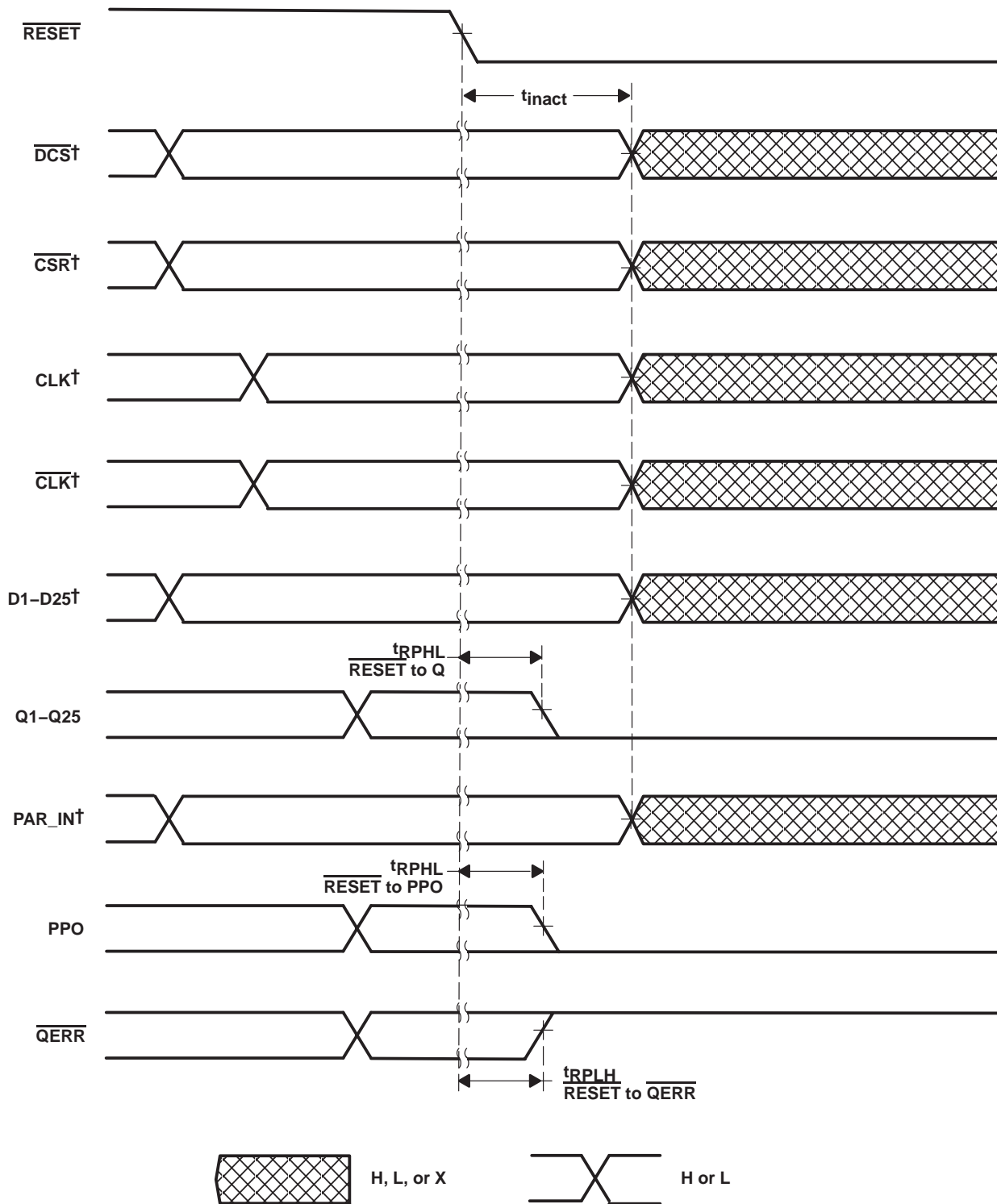
† If the data is clocked in on the n clock pulse, the QERR output signal will be generated on the n + 2 clock pulse, and it will be valid on n + 3 clock pulse. If an error occurs and the QERR output is driven low, it stays latched low for a minimum of two clock cycles or until RESET is driven low.

SN74SSTU32866

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timing diagram for SN74SSTU32866 used as a single device; C0 = 0, C1 = 0
(RESET switches from H to L)



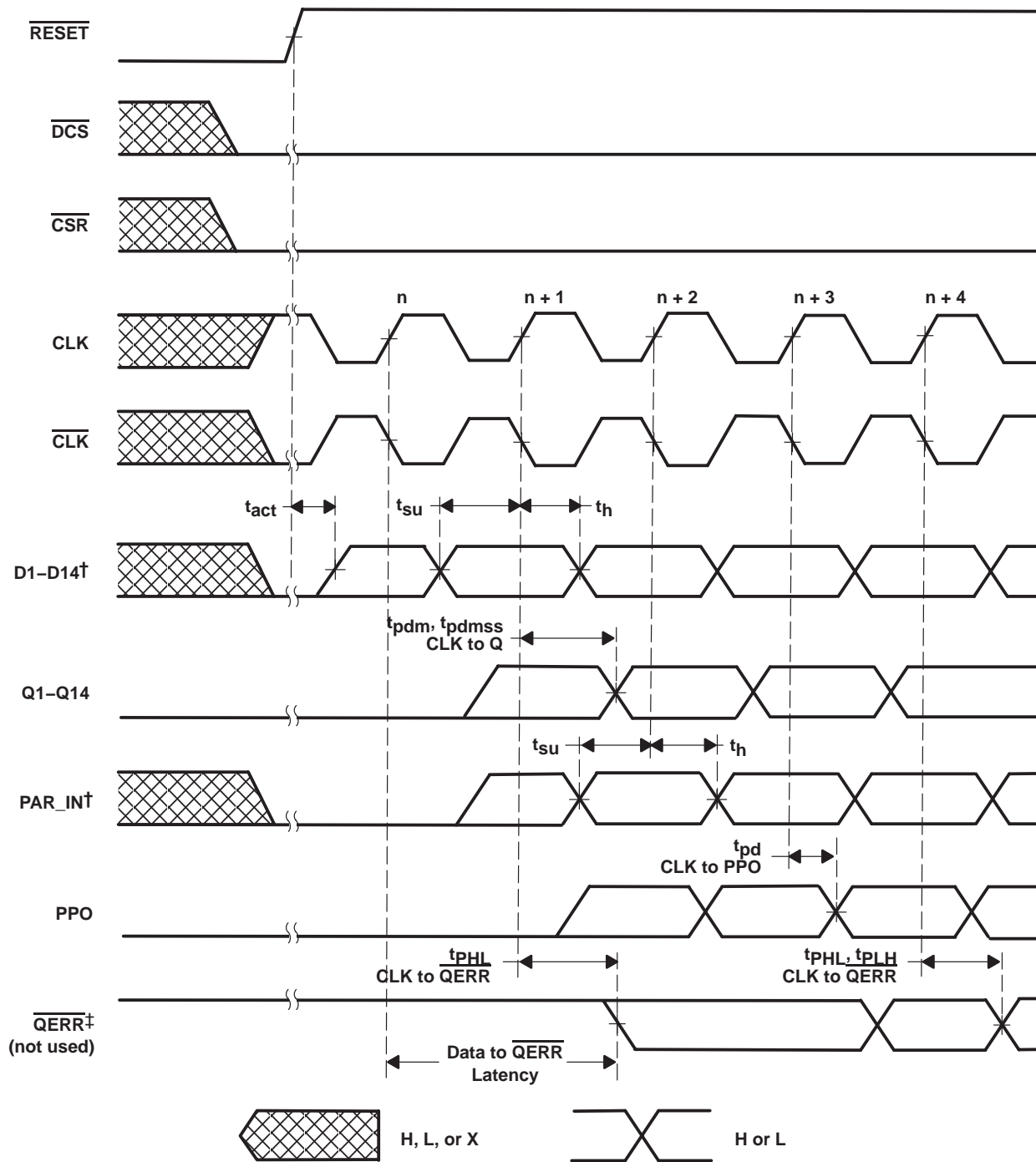
† After $\overline{\text{RESET}}$ is switched from high to low, all data and clock input signals must be held at valid logic levels (not floating) for a minimum time of t_{inact} max.

SN74SSTU32866

25-BIT CONFIGURABLE REGISTERED BUFFER WITH ADDRESS-PARITY TEST

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timing diagram for the first SN74SSTU32866 (1:2 Register-A configuration) device used in pair; C0 = 0, C1 = 1 (RESET switches from L to H)



† After $\overline{\text{RESET}}$ is switched from low to high, all data and PAR_IN input signals must be set and held low for a minimum time of t_{act} max, to avoid false error.

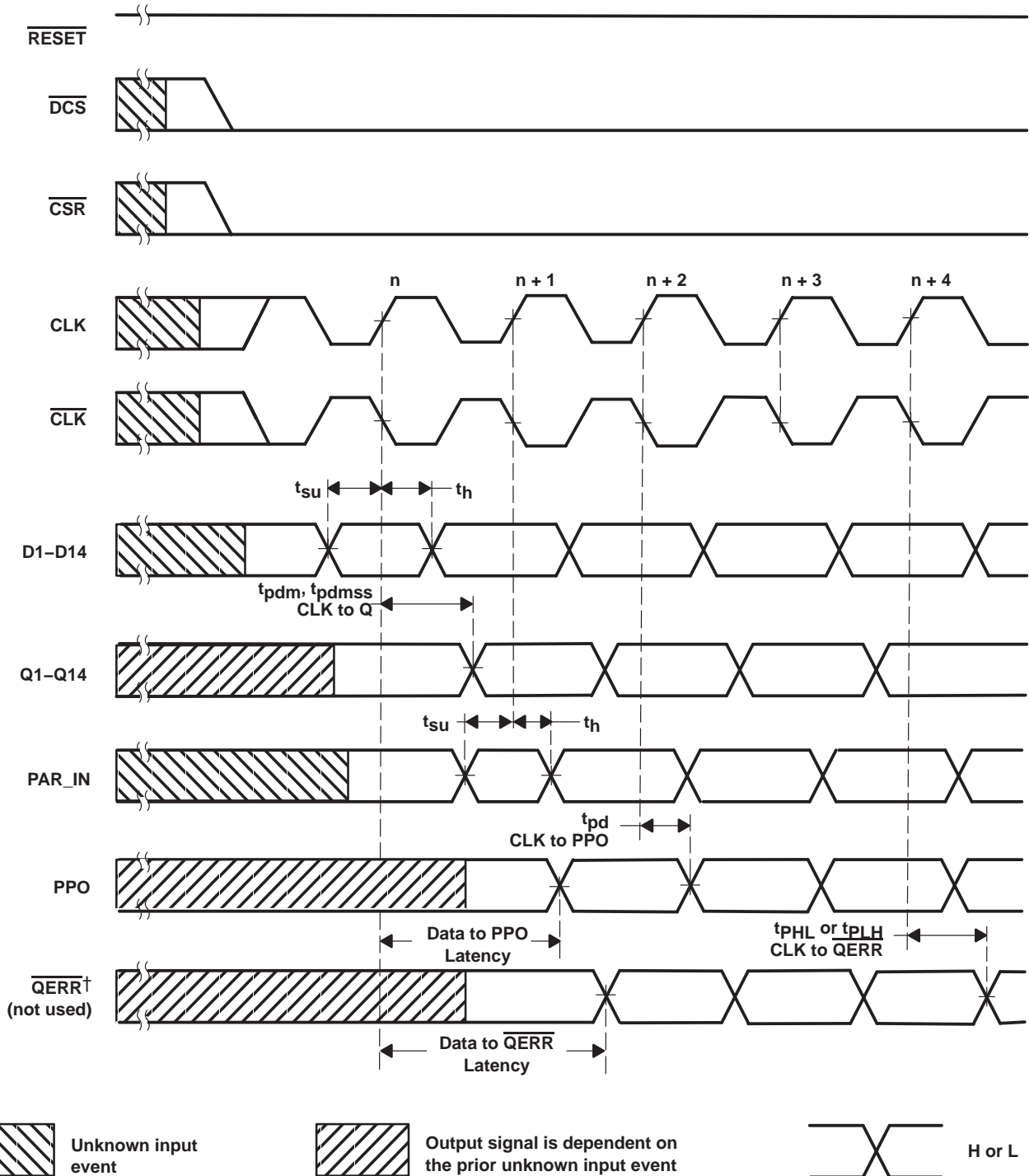
‡ If the data is clocked in on the n clock pulse, the $\overline{\text{QERR}}$ output signal will be generated on the $n + 1$ clock pulse, and it will be valid on the $n + 2$ clock pulse.

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timing diagram for the first SN74SSTU32866 (1:2 Register-A configuration) device used in pair;
C0 = 0, C1 = 1 (RESET = H)



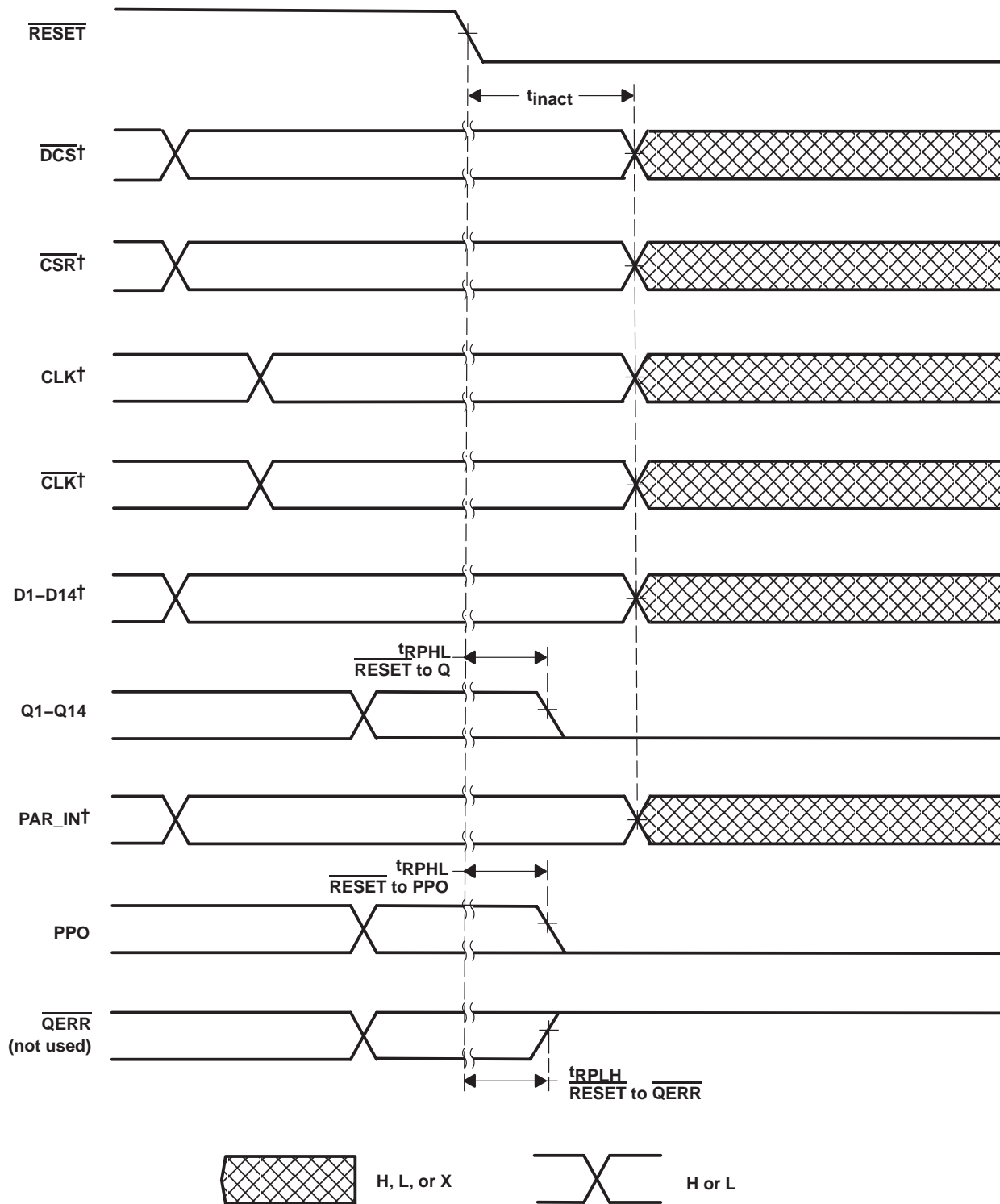
† If the data is clocked in on the n clock pulse, the QERR output signal will be generated on the n + 1 clock pulse, and it will be valid on n + 2 clock pulse. If an error occurs and the QERR output is driven low, it stays latched low for a minimum of two clock cycles or until RESET is driven low.

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25-BIT CONFIGURABLE REGISTERED BUFFER WITH ADDRESS-PARITY TEST

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timing diagram for the first SN74SSTU32866 (1:2 Register-A configuration) device used in pair; C0 = 0, C1 = 1 (RESET switches from H to L)



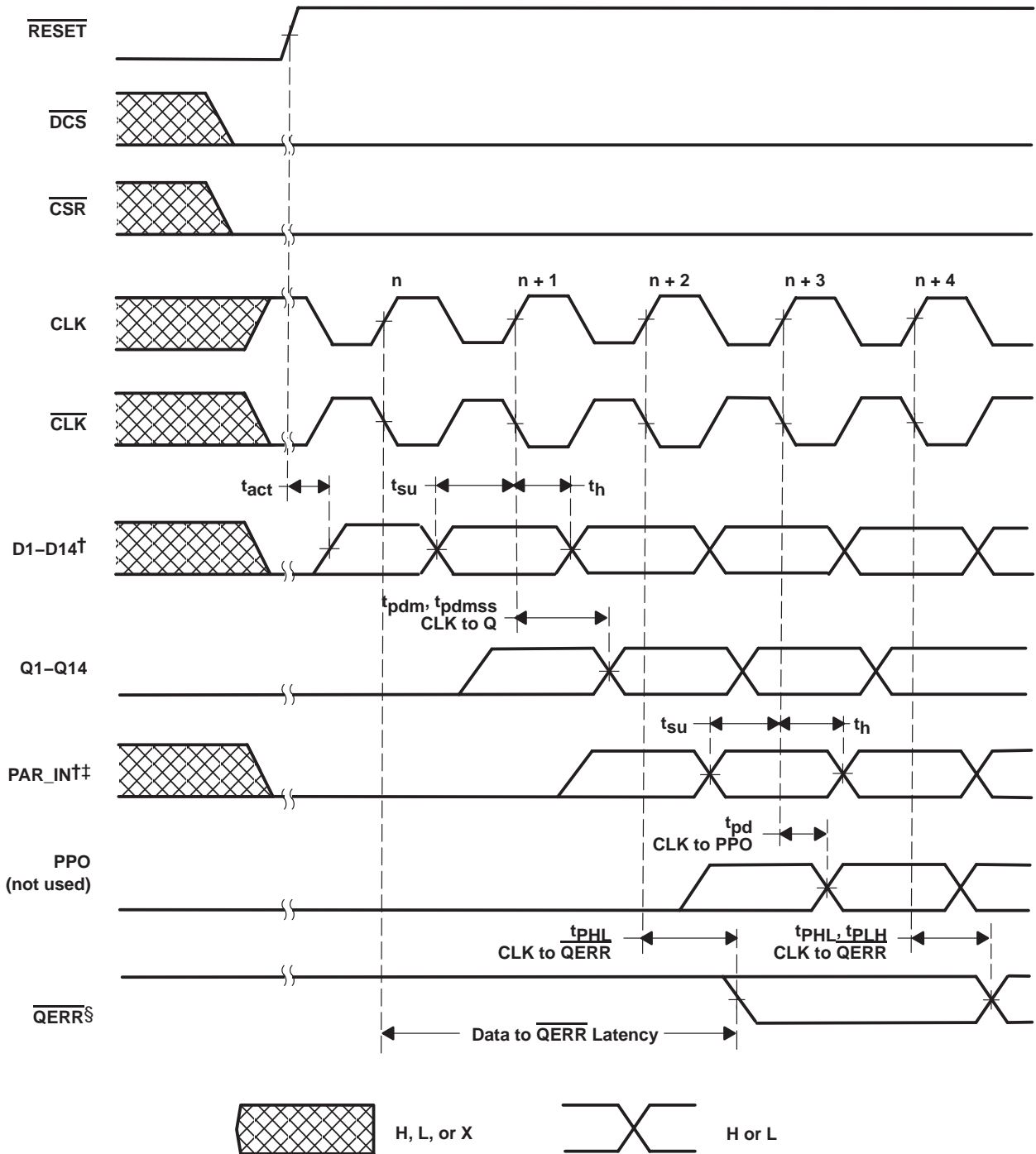
† After $\overline{\text{RESET}}$ is switched from high to low, all data and clock input signals must be held at valid logic levels (not floating) for a minimum time of $t_{\text{inact max}}$.

SN74SSTU32866

25-BIT CONFIGURABLE REGISTERED BUFFER WITH ADDRESS-PARITY TEST

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timing diagram for the second SN74SSTU32866 (1:2 Register-B configuration) device used in pair;
C0 = 1, C1 = 1 (RESET switches from L to H)



† After \overline{RESET} is switched from low to high, all data and $\overline{PAR_IN}$ input signals must be set and held low for a minimum time of t_{act} max, to avoid false error.

‡ $\overline{PAR_IN}$ is driven from PPO of the first SN74SSTU32866 device.

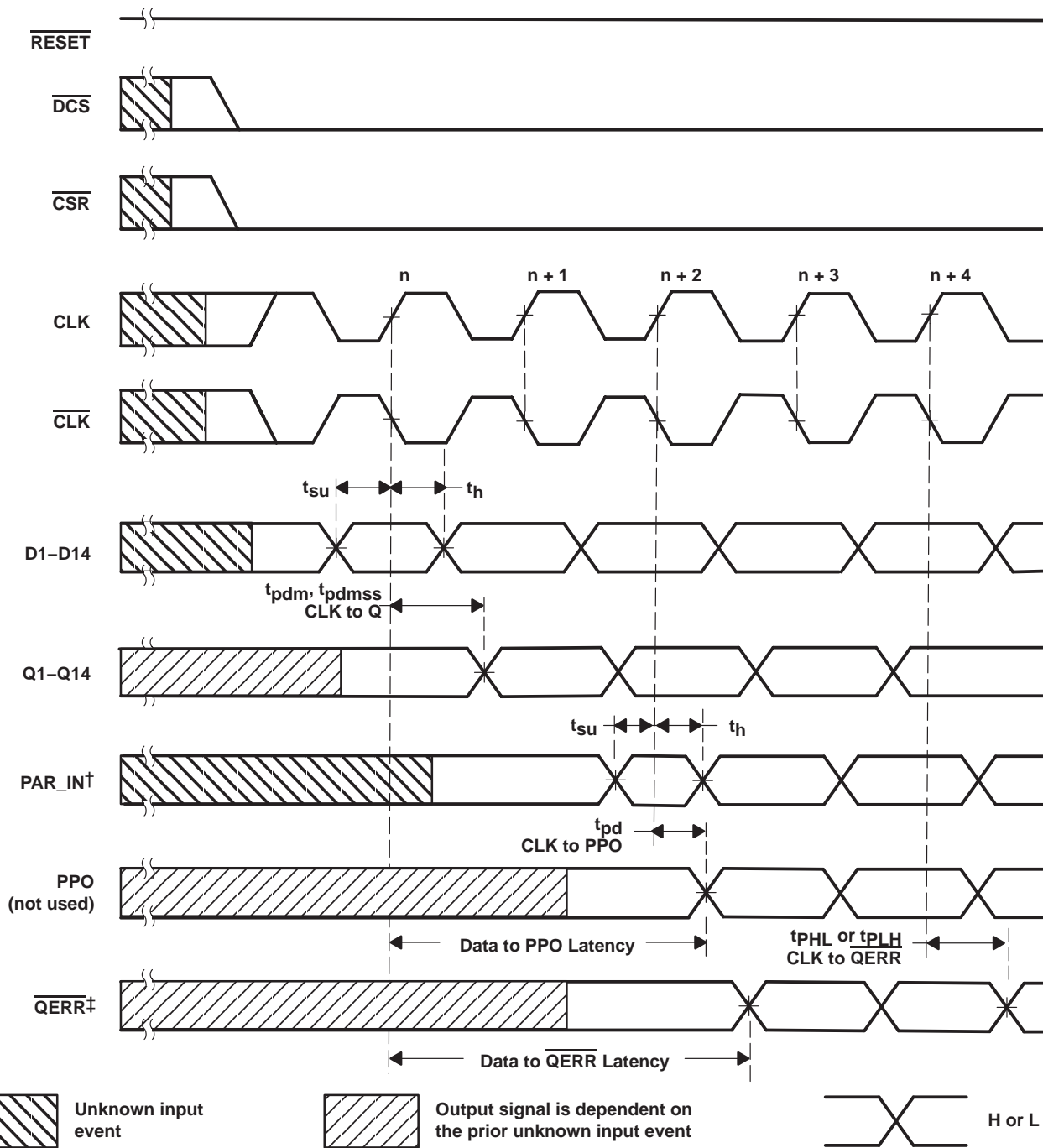
§ If the data is clocked in on the n clock pulse, the \overline{QERR} output signal will be generated on the n + 2 clock pulse, and it will be valid on the n + 3 clock pulse.

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25-BIT CONFIGURABLE REGISTERED BUFFER WITH ADDRESS-PARITY TEST

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timing diagram for the second SN74SSTU32866 (1:2 Register-B configuration) device used in pair; C0 = 1, C1 = 1 (RESET = H)



† PAR_IN is driven from PPO of the first SN74SSTU32866 device.

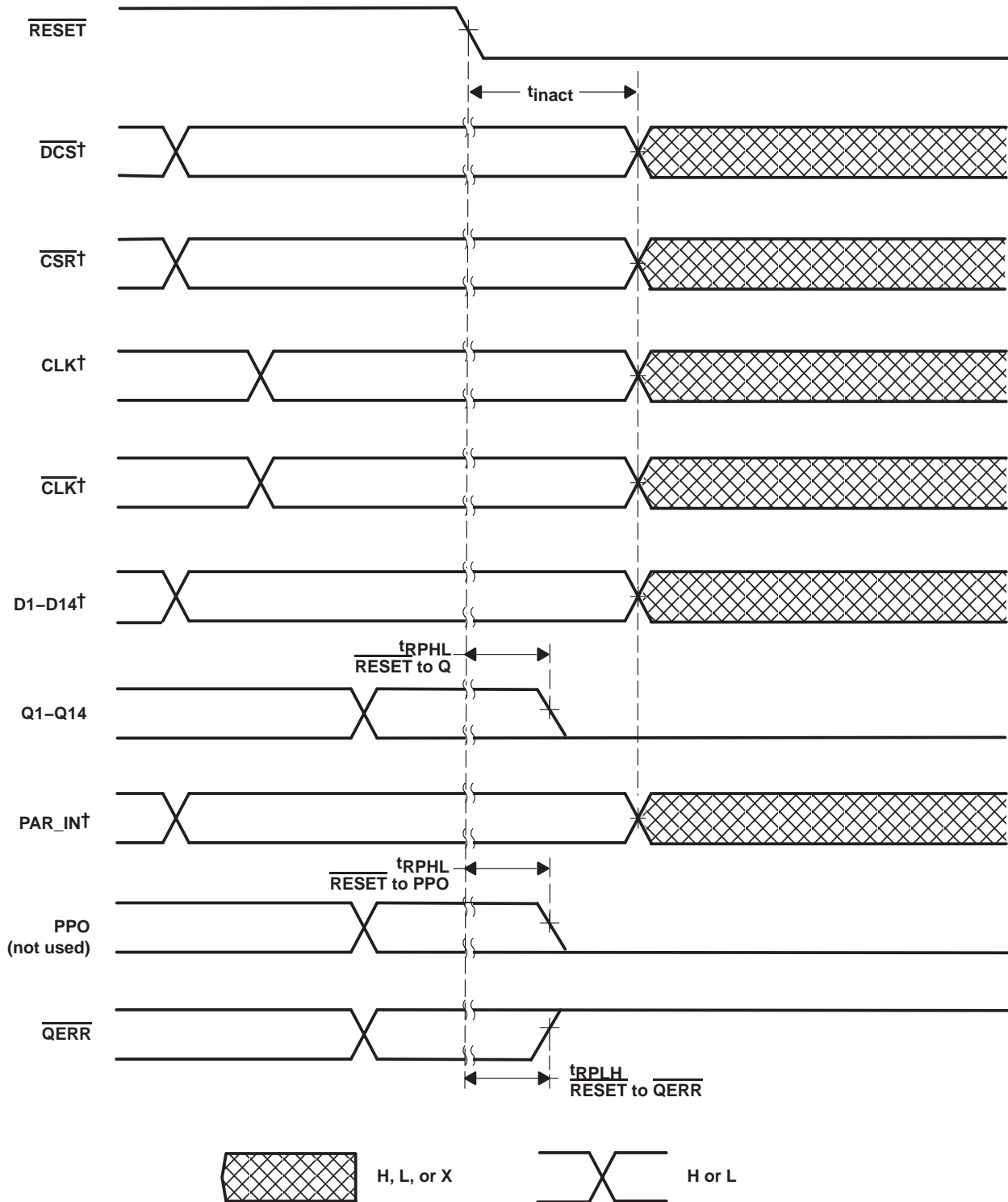
‡ If the data is clocked in on the n clock pulse, the QERR output signal will be generated on the n + 2 clock pulse, and it will be valid on n + 3 clock pulse. If an error occurs and the QERR output is driven low, it stays latched low for a minimum of two clock cycles or until RESET is driven low.

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timing diagram for the second SN74SSTU32866 (1:2 Register-B configuration) device used in pair;
C0 = 1, C1 = 1 (RESET switches from H to L)



† After $\overline{\text{RESET}}$ is switched from high to low, all data and clock input signals must be held at valid logic levels (not floating) for a minimum time of t_{inact} max.

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
SN74SSTU32866ZKER	ACTIVE	LFBGA	ZKE	96		TBD	Call TI	Call TI	0 to 70	SU866	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

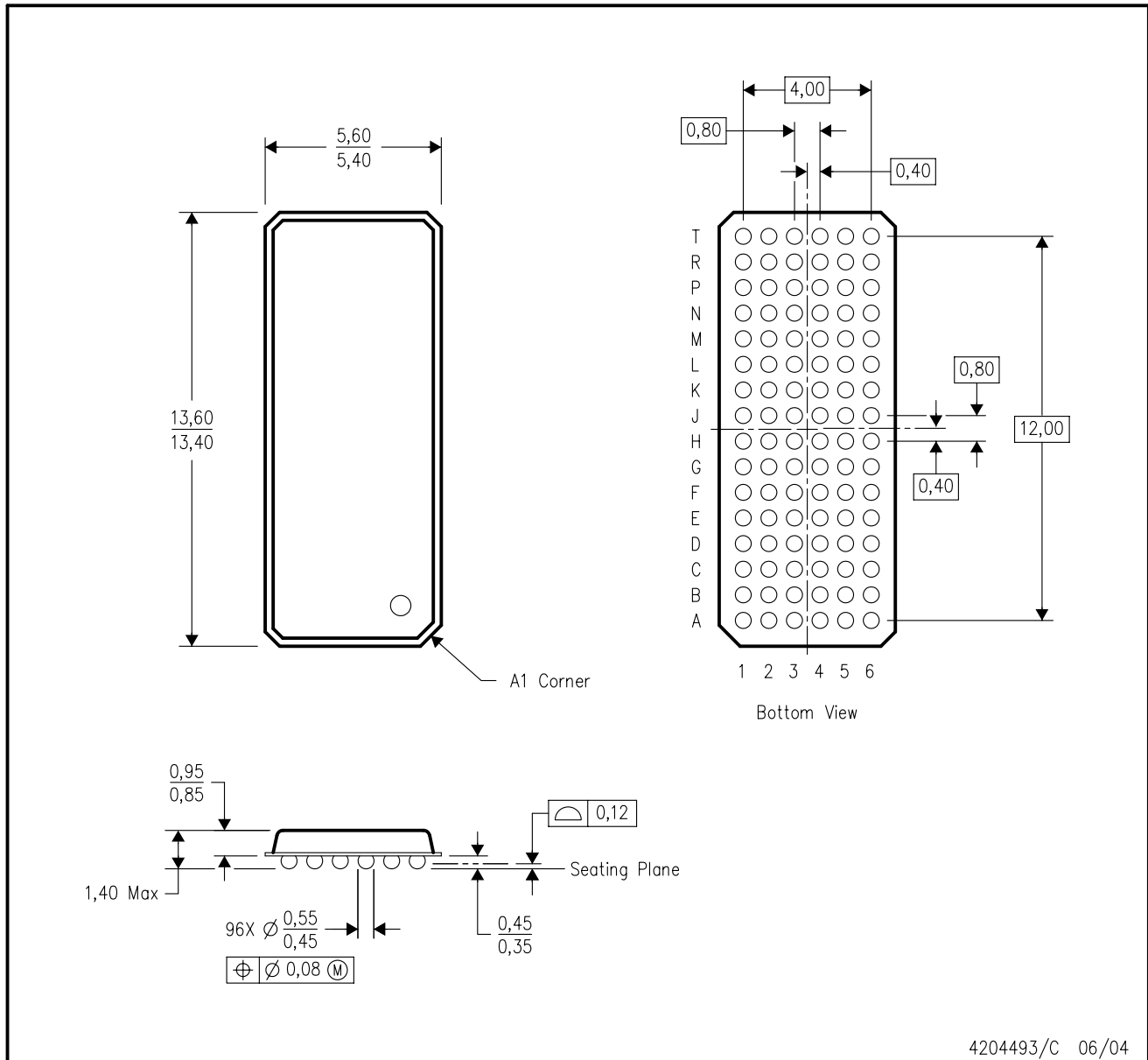
(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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ZKE (R-PBGA-N96)

PLASTIC BALL GRID ARRAY



4204493/C 06/04

- NOTES:
- A. All linear dimensions are in millimeters.
 - B. This drawing is subject to change without notice.
 - C. Falls within JEDEC MO-205 variation CC.
 - D. This package is lead-free. Refer to the 96 GKE package (drawing 4188953) for tin-lead (SnPb).

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