

## DESCRIPTION

The MP20142 is a dual-channel, low noise, low dropout and high PSRR linear regulator with programmable output voltage. The output voltage of MP20142 ranges from 1.2V to 3.3V in 100mV increments and 2.5% accuracy by operating from a +2.5V to +5.5V input. The different output voltage options can be programmed by setting the voltage of P1 and P2 to VDD, GND or floating. There are 9 different voltage settings in one chip. Each output channel can supply up to 200mA of load current.

The MP20142 uses two internal P-channel MOSFETs as the pass devices, which consumes 150µA supply current when both channels are enabled at no load condition. The EN1 and EN2 pins control each output respectively. When both channels shutdown simultaneously, the chip turns off and consumes nearly zero operation current which is suitable for battery-power devices. The MP20142 also features current limiting, output discharge and over temperature protection.

It is available in a TQFN8 (2mm×2mm) package.

## FEATURES

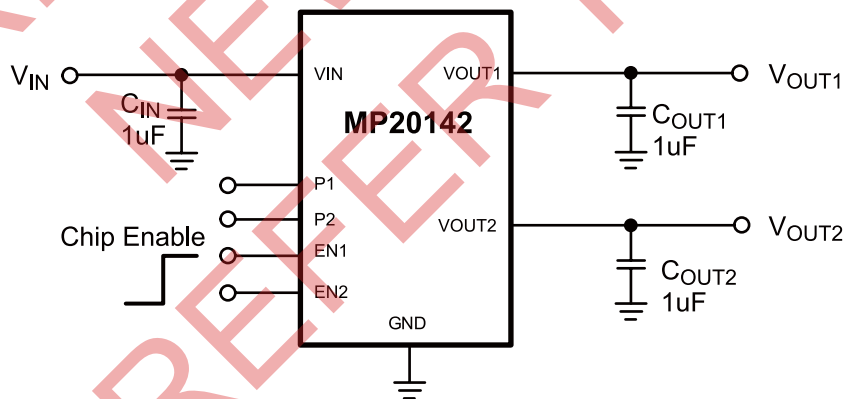
- Programmable Output Voltage
- Up to 200mA Output Current (Each LDO)
- 60mV Dropout at 100mA Load
- Dual Shutdown Pins Control Each Output
- 75dB PSRR at 1kHz
- 7µV<sub>RMS</sub> Low Noise Outputs
- Current Limiting and Thermal Protection
- Short Circuit Protection
- Two LDOs in TQFN8 (2mm×2mm) Package
- Output Load Discharge when Disabled

## APPLICATIONS

- Cellular Phones
- Battery-powered Equipment
- Laptop, Notebook, and Palmtop Computers
- Hand-held Equipment
- Wireless LAN

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## TYPICAL APPLICATION

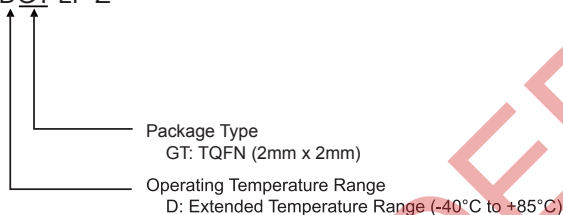


## ORDERING INFORMATION\*

Part Number*	Package	Top Marking	Free Air Temperature (T <sub>A</sub> )
MP20142DGT-LF-Z	TQFN8 (2mm×2mm)	AAY	-40°C to +85°C

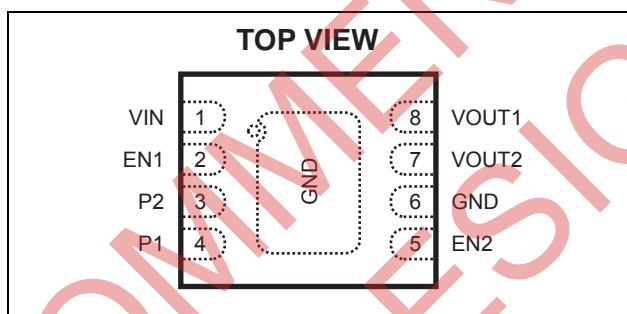
## ORDERING GUIDE\*

MP20142DGT-LF-Z



\* For RoHS Compliant Packaging, add suffix - LF (e.g. MP20142DGT-LF);  
For Tape and Reel, add suffix -Z (e.g. MP20142DGT-LF-Z).

## PACKAGE REFERENCE

ABSOLUTE MAXIMUM RATINGS <sup>(1)</sup>

Supply Input Voltage .....	6V
All Other Pins .....	5.5V
Power Dissipation, PD @ T <sub>A</sub> = 25°C, <sup>(2)</sup>	
TQFN8 .....	1.5W
Storage Temperature Range .....	-65°C to 150°C
Lead Temperature (Soldering, 10sec) .....	260°C

## ESD Susceptibility

HBM (Human Body Mode) .....	2kV
MM (Machine Mode) .....	200V

Recommended Operating Conditions <sup>(3)</sup>

Supply Input Voltage .....	2.5V to 5.5V
Operating Junct. Temp (T <sub>J</sub> ) .....	-40°C to +125°C

Thermal Resistance <sup>(4)</sup>

	$\theta_{JA}$	$\theta_{JC}$
TQFN8 .....	80	16 °C/W

## Notes:

- Exceeding these ratings may damage the device.
- The maximum allowable power dissipation is a function of the maximum junction temperature T<sub>J</sub>(MAX), the junction-to-ambient thermal resistance  $\theta_{JA}$ , and the ambient temperature T<sub>A</sub>. The maximum allowable continuous power dissipation at any ambient temperature is calculated by P<sub>D</sub>(MAX) = (T<sub>J</sub>(MAX) - T<sub>A</sub>) /  $\theta_{JA}$ . Exceeding the maximum allowable power dissipation will cause excessive die temperature, and the regulator will go into thermal shutdown. Internal thermal shutdown circuitry protects the device from permanent damage.
- The device is not guaranteed to function outside of its operating conditions.
- Measured on JESD51-7 4-layer board.

## PROGRAMMABLE OUTPUT VOLTAGE SETTING

P1 State	P2 State	V <sub>OUT1</sub>	V <sub>OUT2</sub>
L	L	1.5V	1.8V
L	H	1.5V	2.8V
L	Open	1.8V	3.0V
H	L	2.6V	3.0V
H	H	2.8V	3.0V
H	Open	2.8V	3.3V
Open	L	3.0V	3.0V
Open	H	3.0V	3.3V
Open	Open	3.3V	3.3V

## ELECTRICAL CHARACTERISTICS

$V_{IN} = (V_{OUT} + 0.4V)$  or  $+2.5V$ , whichever is greater.  $C_{IN} = C_{OUT1} = C_{OUT2} = 1\mu F$ ,  $EN1 = EN2 = V_{IN}$

Typical Value at  $T_A = 25^\circ C$  for each LDO, unless otherwise noted.

Parameter	Symbol	Condition	Min	Typ	Max	Units
Output Voltage Accuracy	$\Delta V_{OUT}$	$I_{LOAD} = 1mA$	-2.5		2.5	%
Maximum Output Current	$I_{MAX}$	Continuous	200			mA
Current Limit	$I_{LIM}$	Short circuit current limit		660		mA
Quiescent Current	$I_G$	No Load (Single Channel)		90		$\mu A$
		No Load (Dual Channel)		150		$\mu A$
Shutdown Supply Current	$I_{GSD}$	$EN1 = EN2 = GND$		0.01	1	$\mu A$
Dropout Voltage <sup>(5)</sup>	$V_{DROP}$	$I_{OUT} = 100mA$		60	120	mV
Line Regulation <sup>(6)</sup>	$\Delta V_{LINE}$	$V_{IN} = (V_{OUT} + 0.4V)$ or $2.5V$ to $5.5V$ , $I_{OUT} = 1mA$	-		0.1	%/V
Load Regulation <sup>(7)</sup>	$\Delta V_{LOAD}$	$V_{OUT} > 1.5V$			1.2	%
		$V_{OUT} \leq 1.5V$			1.8	%
EN Logic High	$V_{IH}$	$V_{IN} = 2.5V$ to $5.5V$	1.6			V
EN Logic Low	$V_{IL}$	$V_{IN} = 2.5V$ to $5.5V$			0.4	V
EN Input Bias Current	$I_{SD}$	$EN = GND$ or $V_{IN}$			100	nA
Thermal Shutdown Temperature	$T_{SD}$			140		$^\circ C$
Thermal Shutdown Hysteresis	$\Delta T_{SD}$			10		$^\circ C$
Output Voltage Noise	$V_{NOISE}$	10Hz to 30kHz, $C_{OUT} = 1\mu F$ , $I_{LOAD} = 1mA$		7		$\mu V_{RMS}$
Output Voltage AC PSRR		100Hz, $C_{OUT} = 1\mu F$ , $I_{LOAD} = 200mA$ (Single Channel loaded)		75		dB
		1kHz, $C_{OUT} = 1\mu F$ , $I_{LOAD} = 200mA$ (Single Channel loaded)		75		dB
Output Voltage Discharge Resistance in Shutdown		$V_{IN} = 5V$ , $EN1 = EN2 = GND$	20		100	$\Omega$
P1/P2 Termination bias for setting High, Low, Open States		$V_{IN} = 2.5V$ to $5.5V$	High state bias current		5	$\mu A$
			Low state bias current	-8		
			Open state		0.7	V

### Notes:

5) Dropout Voltage is defined as the input to output differential when the output voltage drops 100mV below its nominal value.

$$6) \text{ Line Regulation} = \frac{V_{OUT[V_{IN(MAX)}]} - V_{OUT[V_{IN(MIN)}]}}{[V_{IN(MAX)} - V_{IN(MIN)}] \times V_{OUT(NOM)}} \times 100(\% / V)$$

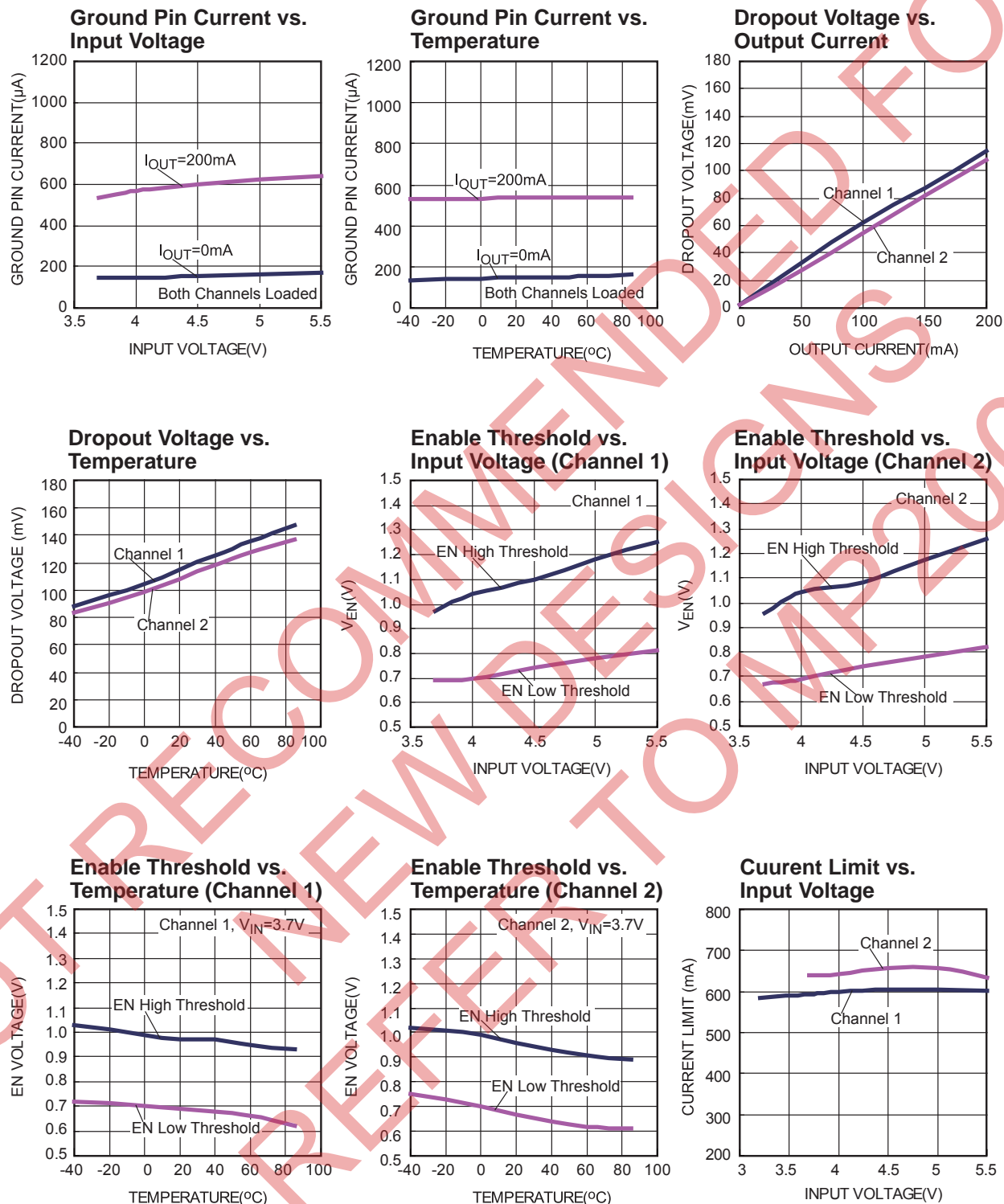
$$7) \text{ Load Regulation} = \frac{V_{OUT[I_{OUT(MAX)}]} - V_{OUT[I_{OUT(MIN)}]}}{V_{OUT(NOM)}} \times 100(\%)$$

## PIN FUNCTIONS

Pin #	Name	Description
1	VIN	Supply Input Pin
2	EN1	Channel 1 Enable (Active High). Do Not Float This Pin.
3	P2	Programming input 2. The state of P2 selects one of nine output voltage options.
4	P1	Programming input 1. The state of P1 selects one of nine output voltage options.
5	EN2	Channel 2 Enable (Active High). Do Not Float This Pin.
6	GND	Common Ground
7	VOUT2	Channel 2 Output Voltage
8	VOUT1	Channel 1 Output Voltage

## TYPICAL PERFORMANCE CHARACTERISTICS

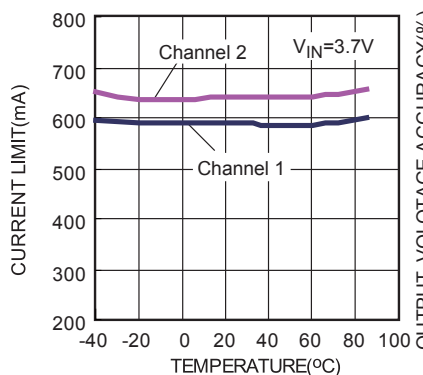
$V_{IN} = 3.7V$ , P1 High, P2 Open ( $V_{OUT1}=2.8V$ ,  $V_{OUT2}=3.3V$ ),  $C_{IN} = C_{OUT1} = C_{OUT2}=1\mu F$ ,  $EN1=EN2= V_{IN}$   
 $T_A = +25^{\circ}C$ , unless otherwise noted.



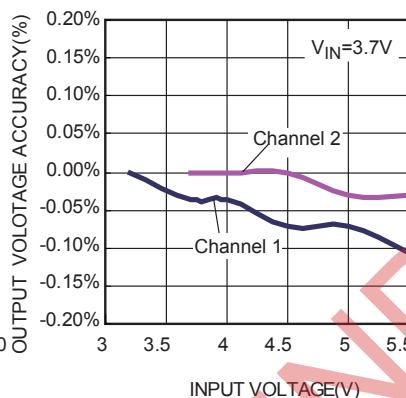
## TYPICAL PERFORMANCE CHARACTERISTICS (continued)

$V_{IN} = 3.7V$ , P1 High, P2 Open ( $V_{OUT1}=2.8V$ ,  $V_{OUT2}=3.3V$ ),  $C_{IN} = C_{OUT1} = C_{OUT2} = 1\mu F$ ,  $EN1=EN2= V_{IN}$   
 $T_A = +25^{\circ}C$ , unless otherwise noted.

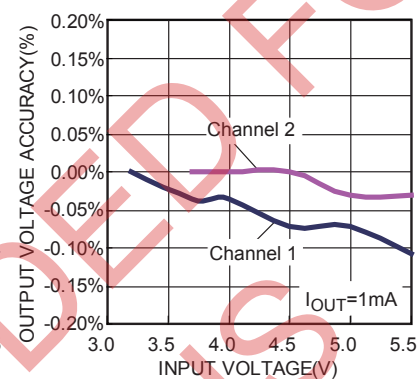
**Current Limit vs. Temperature**



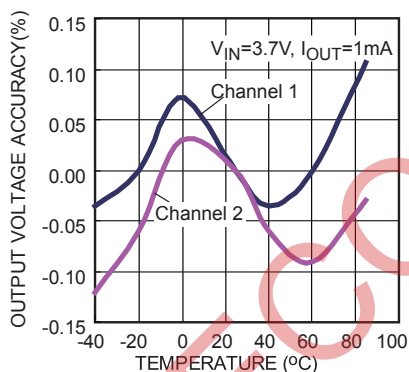
**Load Regulation**



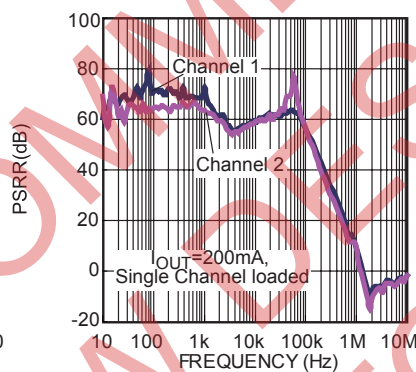
**Line Regulation**



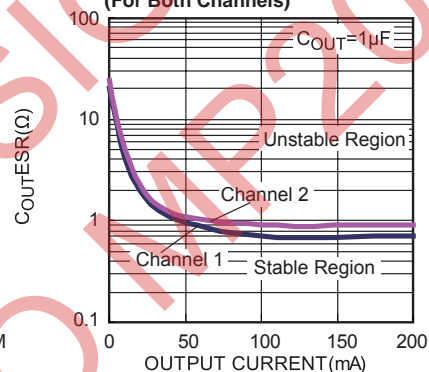
**Output Voltage Accuracy**



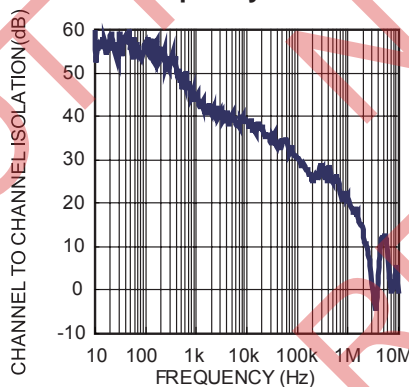
**PSRR**



**$C_{OUT}$  ESR Stability vs. Output Current  
(For Both Channels)**



**Channel to Channel Isolation vs. Frequency**





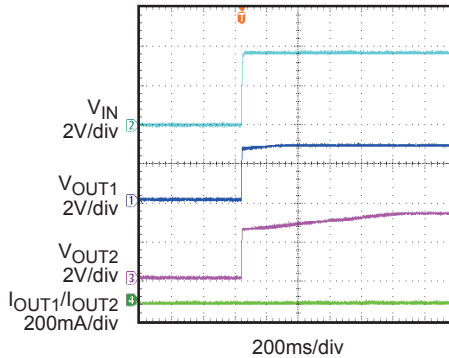
## TYPICAL PERFORMANCE CHARACTERISTICS (continued)

$V_{IN} = 3.7V$ , P1 High, P2 Open ( $V_{OUT1}=2.8V$ ,  $V_{OUT2}=3.3V$ ),  $C_{IN} = C_{OUT1} = C_{OUT2} = 1\mu F$ ,  $EN1=EN2=V_{IN}$

$T_A = +25^\circ C$ , unless otherwise noted.

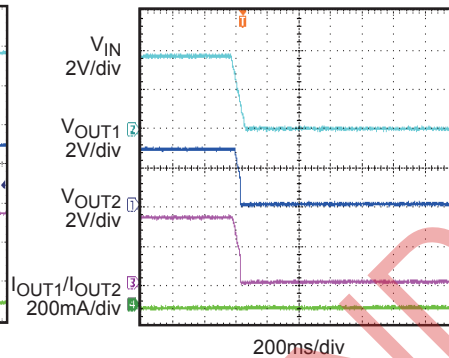
### Input Power Up

$V_{IN}=3.7V$ , No Load



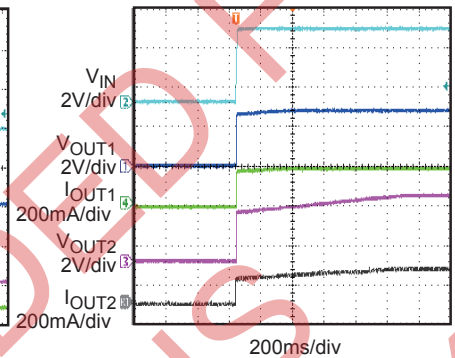
### Input Power Down

$V_{IN}=3.7V$ , No Load



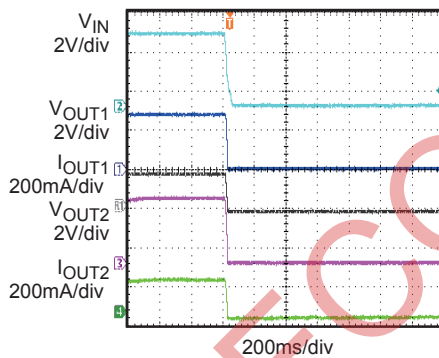
### Input Power Up

$V_{IN}=3.7V$ ,  $I_{OUT1}=I_{OUT2}=200mA$  with Resistive Load



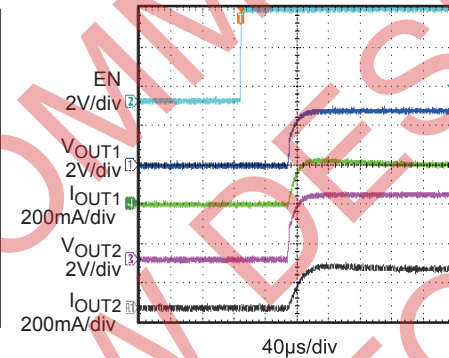
### Input Power Down

$V_{IN}=3.7V$ ,  $I_{OUT1}=I_{OUT2}=200mA$  with Resistive Load



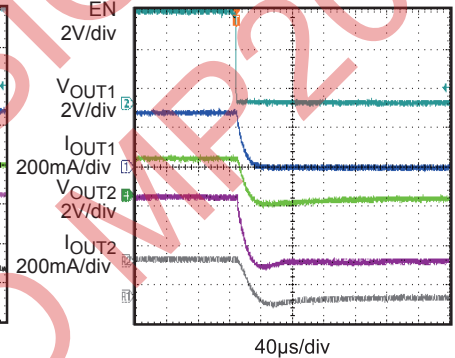
### EN Start Up

$V_{IN}=3.7V$ ,  $I_{OUT1}=I_{OUT2}=200mA$  with Resistive Load



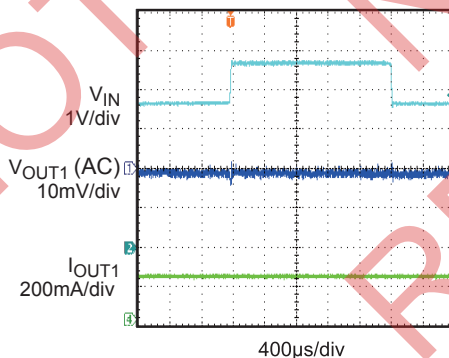
### EN Shut Down

$V_{IN}=3.7V$ ,  $I_{OUT1}=I_{OUT2}=200mA$  with Resistive Load



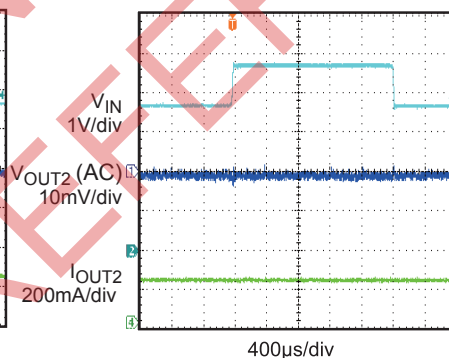
### Line Transient (Ch1)

$V_{IN}=3.7V$  to  $4.7V$ ,  $I_{OUT1}=200mA$  with Resistive Load



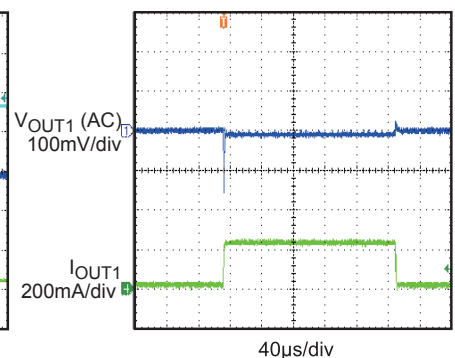
### Line Transient (Ch2)

$V_{IN}=3.7V$  to  $4.7V$ ,  $I_{OUT2}=200mA$  with Resistive Load



### Load Transient (Ch1)

$V_{IN}=3.7V$ ,  $I_{OUT1}=1mA$  to  $200mA$  with Resistive Load



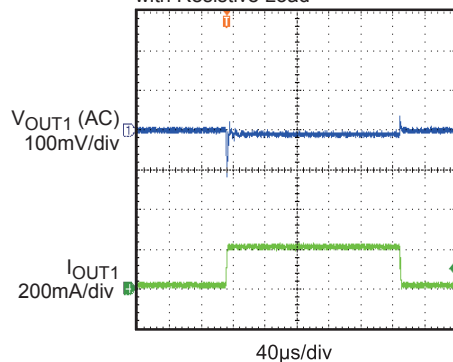


## TYPICAL PERFORMANCE CHARACTERISTICS *(continued)*

$V_{IN} = 3.7V$ , P1 High, P2 Open ( $V_{OUT1}=2.8V$ ,  $V_{OUT2}=3.3V$ ),  $C_{IN} = C_{OUT1} = C_{OUT2} = 1\mu F$ ,  $EN1=EN2= V_{IN}$   
 $T_A = +25^{\circ}C$ , unless otherwise noted.

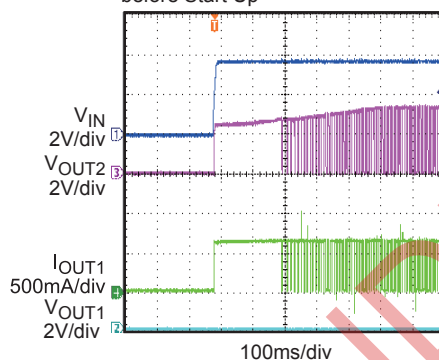
### Load Transient (Ch2)

$V_{IN}=3.7V$ ,  $I_{OUT2}=1mA$  to  $200mA$   
 with Resistive Load



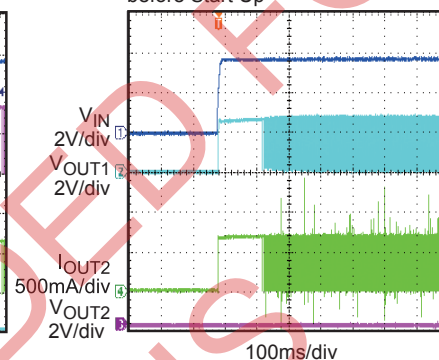
### OCP Start Up

$V_{IN}=3.7V$ , Short  $V_{OUT1}$  to GND  
 before Start Up



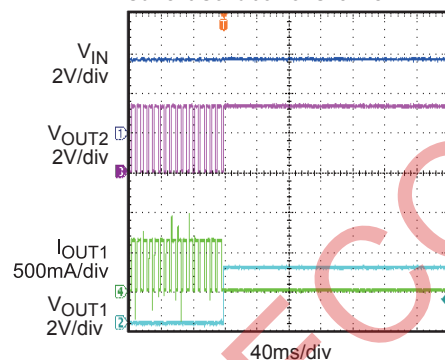
### OCP Start Up

$V_{IN}=3.7V$ , Short  $V_{OUT2}$  to GND  
 before Start Up



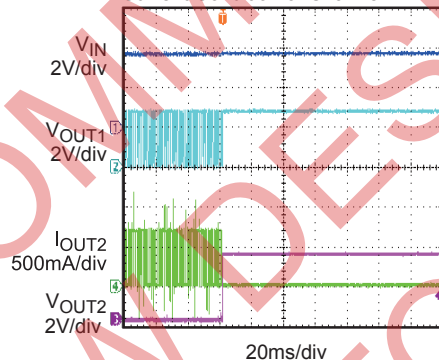
### OCP Recovery

$V_{IN}=3.7V$ , Remove the Over  
 Current Condition of Channel 1

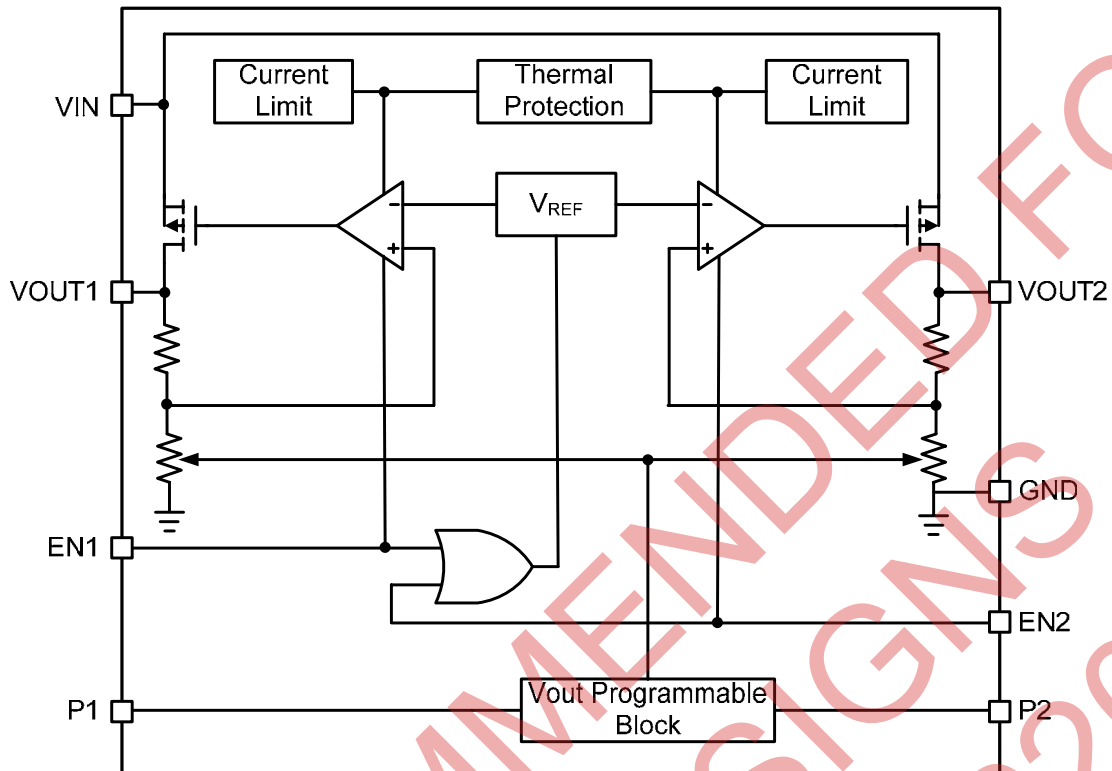


### OCP Recovery

$V_{IN}=3.7V$ , Remove the Over  
 Current Condition of Channel 2



## FUNCTION BLOCK DIAGRAM



### Figure1—Function Block Diagram

## OPERATION

The MP20142 integrates two low noise, low dropout, low quiescent current and high PSRR linear regulators. It is intended for use in devices that require very low voltage, low quiescent current power such as wireless LAN, battery-powered equipment and hand-held equipment. Output voltages are optional ranging from 1.2V to 3.3V, and each channel can supply current up to 200mA. The part uses internal P-channel MOSFETs as the pass elements and features internal thermal shutdown and internal current limit.

### Dropout Voltage

Dropout voltage is the minimum input to output differential voltage required for the regulator to maintain an output voltage within 100mV of its nominal value. Because the P-channel MOSFETs pass element behaves as a low-value resistor, the dropout voltage of MP20142 is very low.

### Shutdown

The MP20142 has two EN pins to control each channel respectively. And each channel can be switched ON or OFF by a logic input at the EN pin. A high voltage at this pin will turn the device on. When the EN pin is low, the regulator output is off. The EN pin should be tied to VIN to keep the regulator output always on if the application does not require the shutdown feature. Do not float the EN pins.

### Output Discharge

The part involves a discharge function that provides a resistive discharge path for the external output capacitor. The function will be active when the part is disabled and it will be done in a very limited time.

### Current Limit

The MP20142 includes two independent current limit structures which monitor and control each P-channel MOSFET's gate voltage limiting the guaranteed maximum output current to 660mA.

### Thermal Protection

Thermal protection turns off the P-channel MOSFETs when the junction temperature exceeds +140°C, allowing the IC to cool. When the IC's junction temperature drops by 10°C, the P-channel MOSFETs will be turned on again. Thermal protection limits total power dissipation in the MP20142. For reliable operation, the junction temperature should be limited to 125 °C maximum.

### Internal P-Channel Pass Device

The MP20142 features dual typical 0.6Ω P-channel MOSFET pass devices. It provides several advantages over similar designs using PNP pass transistors, including longer battery life.

## APPLICATION INFORMATION

### Output Voltage

The output voltage of MP20142 is programmable with setting different states of P1 PIN and P2 PIN. The setting rule is as the following table:

**Table 1 Programmable Output Voltage Setting**

P1 State	P2 State	V <sub>OUT1</sub>	V <sub>OUT2</sub>
L	L	1.5V	1.8V
L	H	1.5V	2.8V
L	Open	1.8V	3.0V
H	L	2.6V	3.0V
H	H	2.8V	3.0V
H	Open	2.8V	3.3V
Open	L	3.0V	3.0V
Open	H	3.0V	3.3V
Open	Open	3.3V	3.3V

For example, when P1 and P2 are both open, output voltage of channel 1 is 3.3V, and the output voltage of channel 2 is 3.3V.

### Power Dissipation

The power dissipation for any package depends on the thermal resistance of the case and circuit board, the temperature difference between the junction and ambient air, and the rate of airflow. The power dissipation across the device can be represented by the equation:

$$P = (V_{IN} - V_{OUT}) \times I_{OUT}$$

The allowable power dissipation can be calculated using the following equation:

$$P_{(MAX)} = (T_{Junction} - T_{Ambient}) / \theta_{JA}$$

Where  $(T_{Junction} - T_{Ambient})$  is the temperature difference between the junction and the surrounding environment,  $\theta_{JA}$  is the thermal resistance from the junction to the ambient environment. Connect the GND pin of MP20142 to ground using a large pad or ground plane helps to channel heat away.

### Input Capacitor Selection

Using a capacitor whose value is  $>0.47\mu F$  on the MP20142 input and the amount of capacitance can be increased without limit.

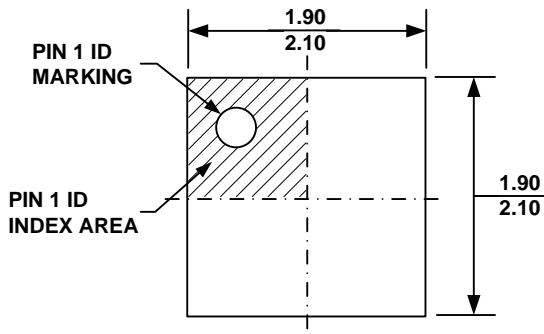
Input capacitor of larger value will help improve line transient response with the drawback of increased size. Ceramic capacitors are preferred, but tantalum capacitors may also sufficient.

### Output Capacitor Selection

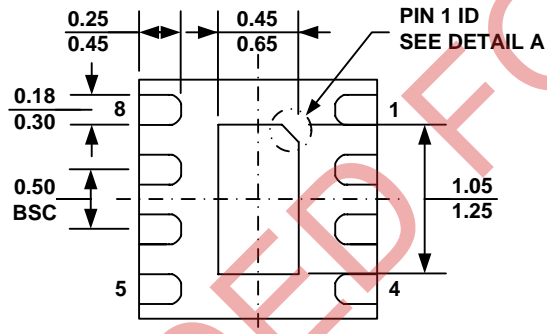
The MP20142 is designed specifically to work with very low ESR ceramic output capacitor in space-saving and performance consideration. A ceramic capacitor in the range of  $0.47\mu F$  and  $10\mu F$ , and with ESR lower than  $1\Omega$  is suitable for the MP20142 application circuit. Output capacitor of larger values will help to improve load transient response and reduce noise with the drawback of increased size.

## PACKAGE INFORMATION

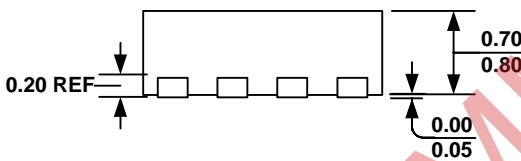
### TQFN8 (2mm×2mm)



**TOP VIEW**



**BOTTOM VIEW**



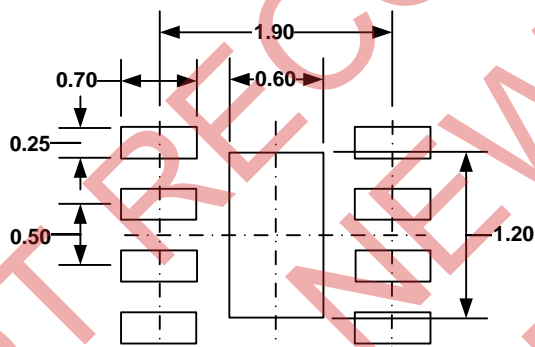
**SIDE VIEW**

**PIN 1 ID OPTION A**  
0.30x45° TYP.

**PIN 1 ID OPTION B**  
R0.20 TYP.



**DETAIL A**



**RECOMMENDED LAND PATTERN**

#### **NOTE:**

- 1) ALL DIMENSIONS ARE IN MILLIMETERS.
- 2) EXPOSED PADDLE SIZE DOES NOT INCLUDE MOLD FLASH.
- 3) LEAD COPLANARITY SHALL BE 0.10 MILLIMETER MAX.
- 4) DRAWING CONFORMS TO JEDEC MO-229, VARIATION WCCD-3.
- 5) DRAWING IS NOT TO SCALE.

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